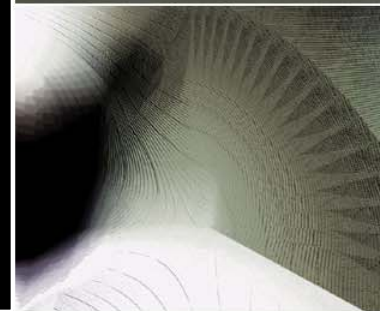


IMECE2005-81600

2005 ASME International Mechanical Engineering Congress and Exposition  
November 5-11, 2005, Orlando, Florida USA

# ANALOG AND LABVIEW- BASED CONTROL OF A MAGLEV SYSTEM WITH NI-ELVIS



## **RISHABH SINHA**

Cooling Design Center  
Large Power Systems Division  
Caterpillar Inc.  
Mossville, IL 61552 USA  
sinha\_rishabh@cat.com

## **MARK NAGURKA**

Department of Mechanical &  
Industrial Engineering  
Marquette University  
Milwaukee, WI 53201 USA  
mark.nagurka@marquette.edu

# Marquette University



# Outline of Presentation

- Objectives
- Background
- Maglev Testbed
- Control Strategies
- Simulation Studies
- Experimental Studies
- Summary
- Future Work

# Objectives

Using a low cost maglev system for mechatronics education:

- o Develop a linearized model
- o Investigate classical linear controllers
- o Implement controllers using analog circuits on NI-ELVIS
- o “Implement” controllers using LabVIEW
- o Compare performance of controllers

# Outline of Presentation

- Objectives
- **Background**
- Maglev Testbed
- Control Strategies
- Simulation Studies
- Experimental Studies
- Summary
- Future Work

# What is Maglev?

- o Maglev = **M**agnetic **L**evitation
- o Levitate objects by electromagnetic forces to cancel effect of gravity
- o Established technology :
  - o high-speed maglev vehicles
  - o maglev bearings
  - o vibration isolation systems, etc.

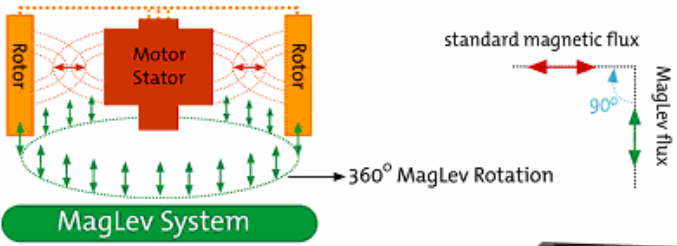
# Examples



**MAGLeV**  
Fan & Blowers  
by SUNON

## MagLeV. Brand Fans

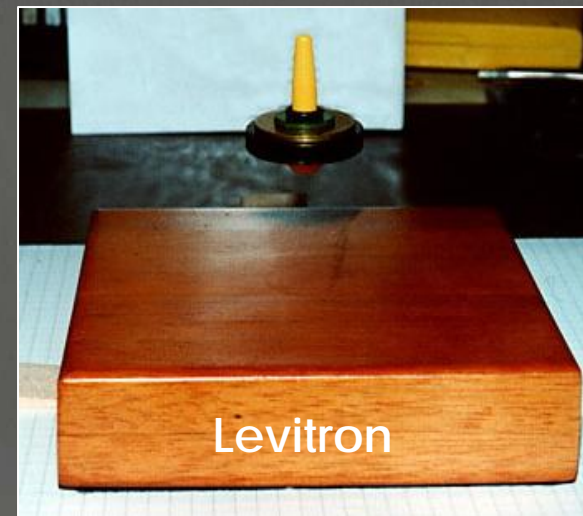

MagLeV. = Standard magnetic flux + MagLeV flux



standard magnetic flux  
90°  
MagLeV flux  
360° MagLeV Rotation  
MagLeV System

**MagLeV. Law**

1. The MagLeV system creates 360° attraction on the rotor, which results in stable rotation.
2. MagLeV flux acts perpendicular to the standard magnetic flux.



# Control Strategies

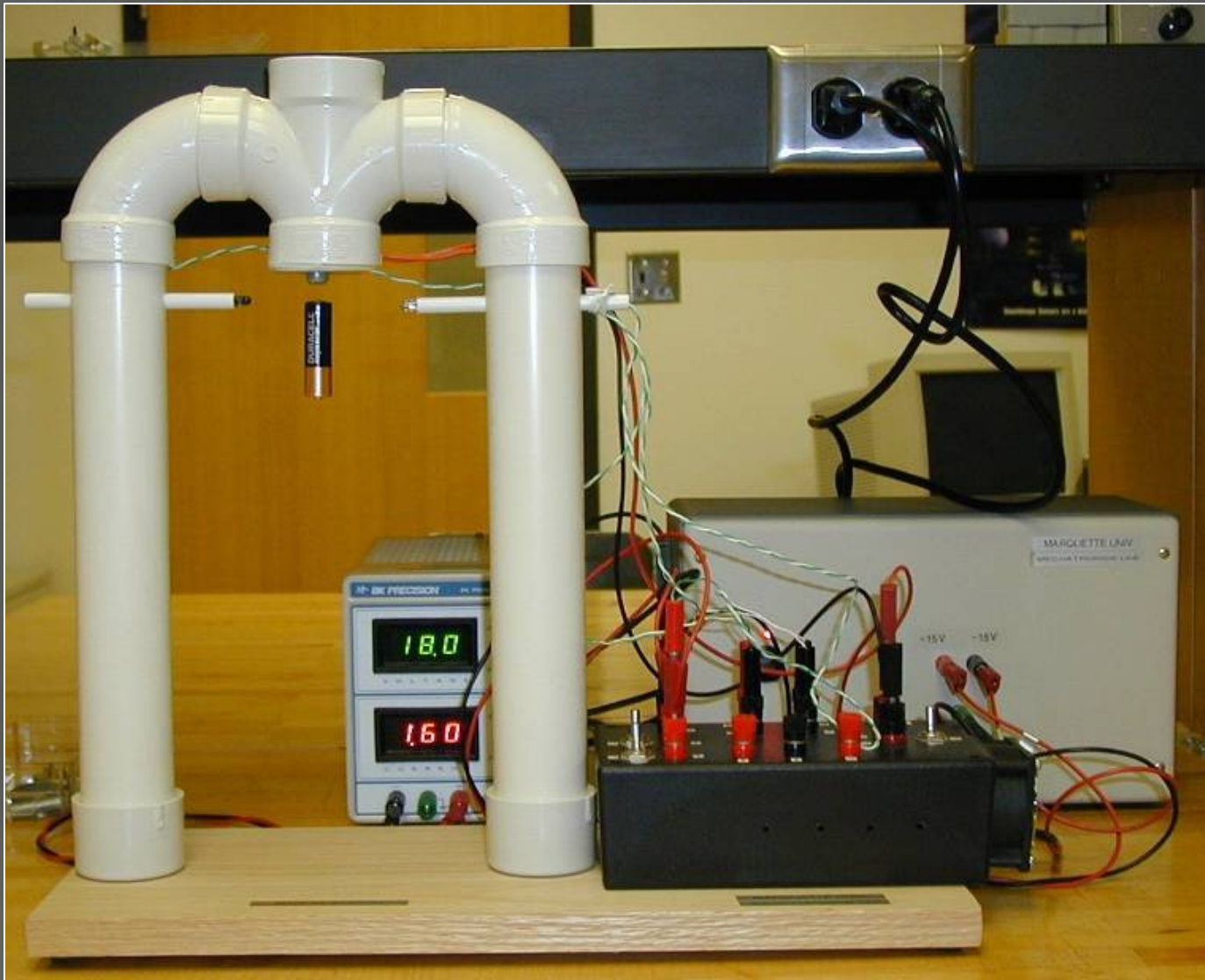
- **Linear Control**
  - Classical PID
  - PID with gain scheduling
  - Phase-lead, Phase-lag
  - LQR, LQE, LQG,  $H_\infty$ ,  $\mu$ -synthesis
- **Nonlinear Control**
  - On-off
  - PWM
  - Fuzzy-logic
  - Neural-network control
  - Feedback linearization
  - Adaptive control
  - Backstepping theory



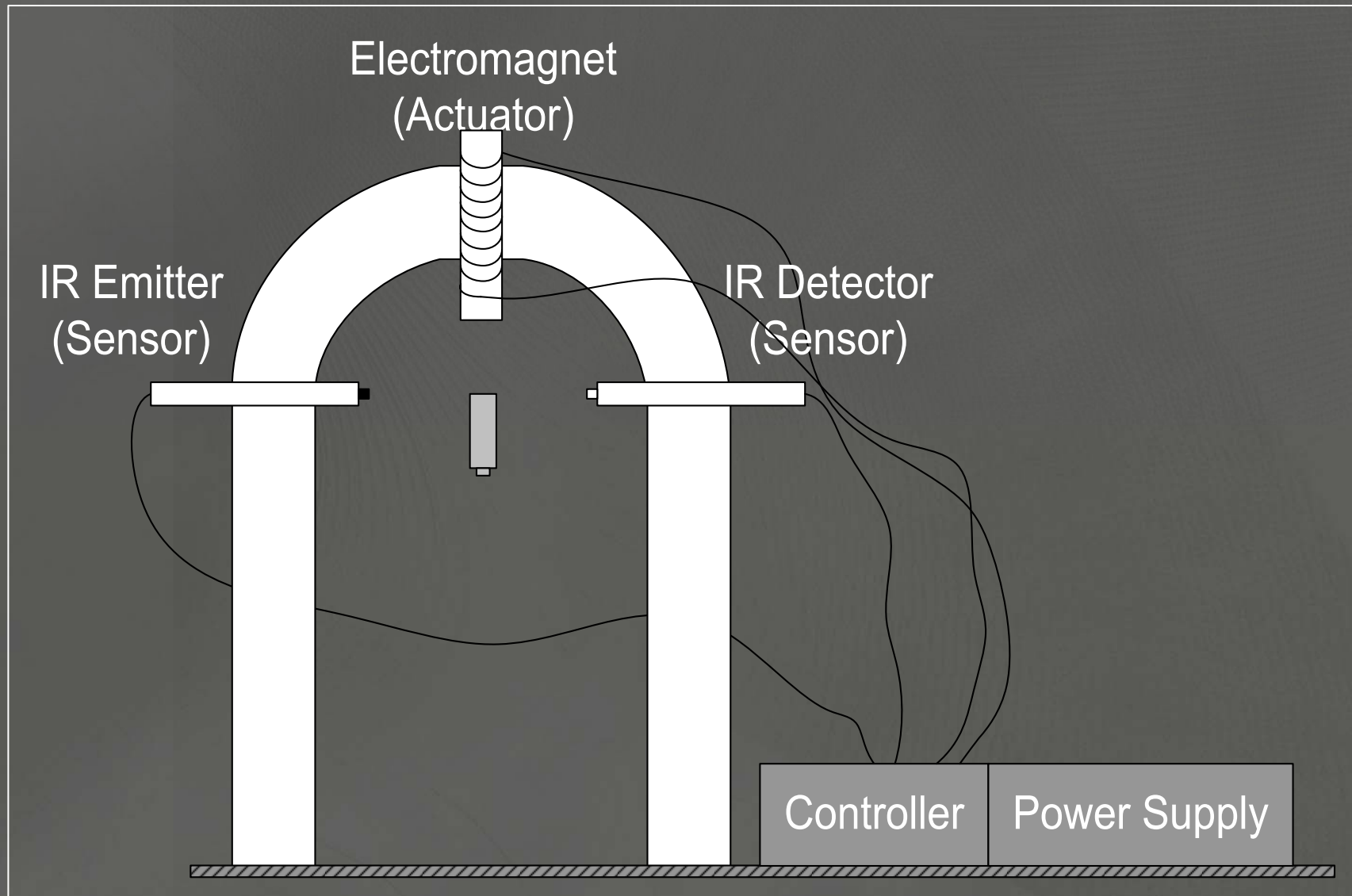
# Outline of Presentation

- Objectives
- Background
- **Maglev Testbed**
- Control Strategies
- Simulation Studies
- Experimental Studies
- Summary
- Future Work

# Physical Testbed

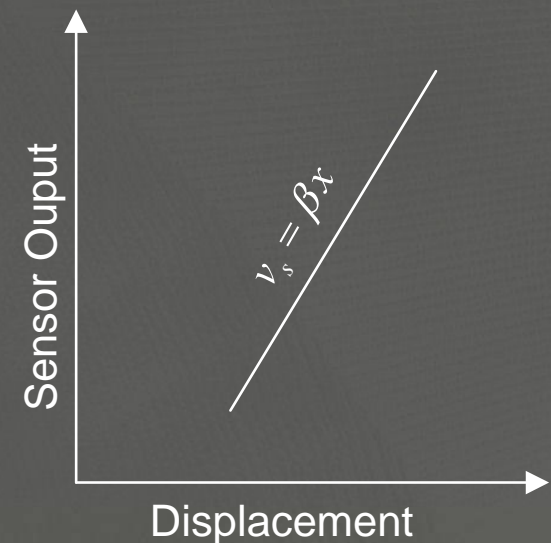


# Testbed Pictorial

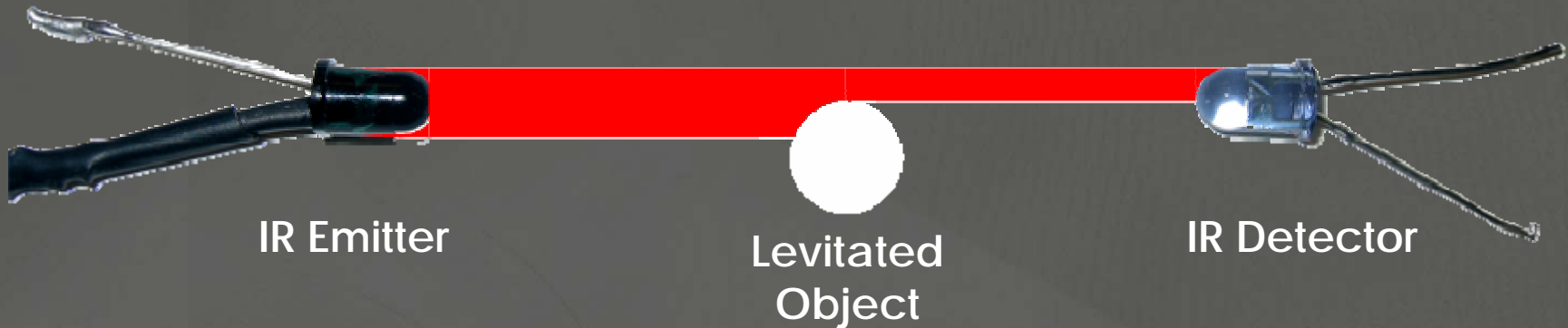


# Infrared (IR) Sensor

- Infrared emitter and detector pair
- Acts as a variable resistor
- Levitated object blocks path, changes light intensity
- Linear behavior in operating region

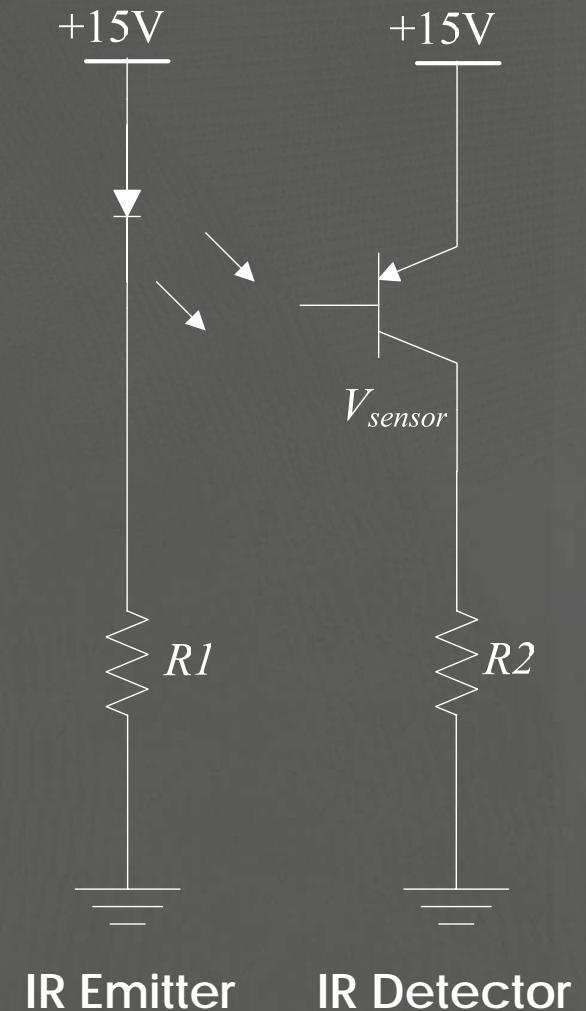


$$v_s = \beta x$$



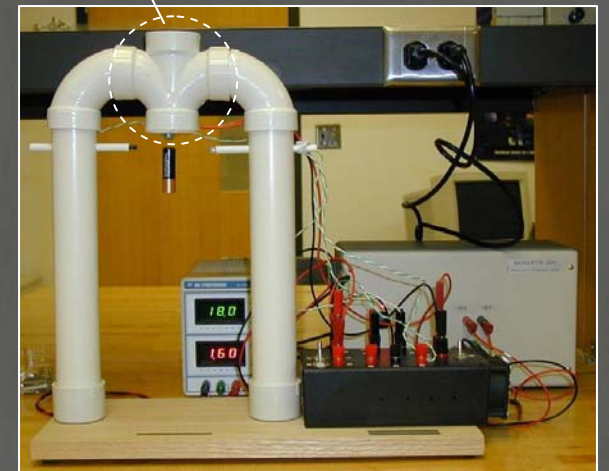
# Sensor and Controls Logic

- The emitter generates constant light intensity.
- The detector signal is amplified and compared with a reference voltage.
- Difference of signals used to adjust current to electromagnet.
  - If the levitated object is too close to electromagnet (detected IR signal too small), the current is reduced.
  - If the levitated object is too far (detected IR signal too large), the current is increased.



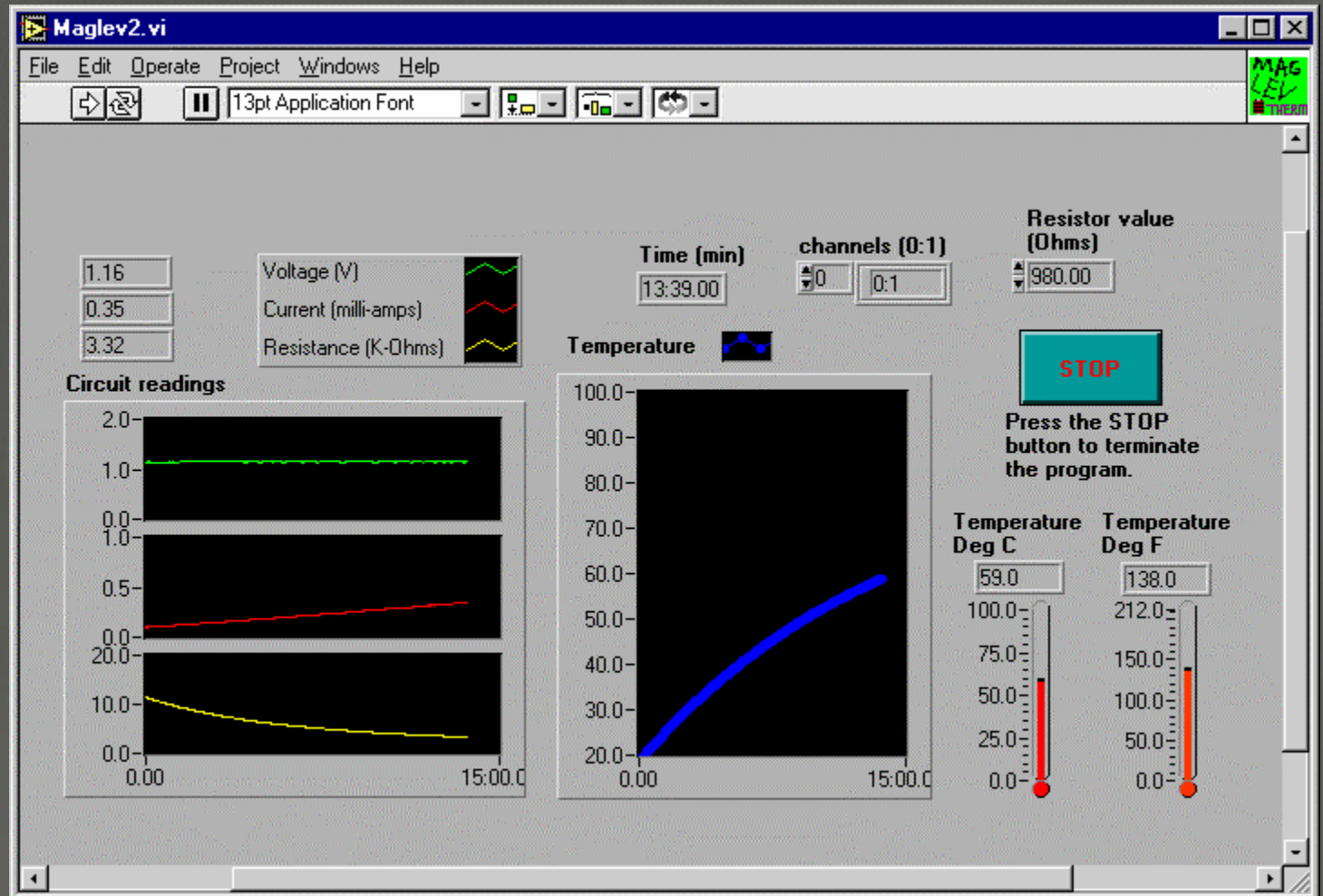
# Actuator

- Actuator is electro-magnetic coil with steel bolt core
- Electromagnet obtained commercially
- FEA performed to check for magnetic saturation
- Power supply: 18 V / 3 A



# Maglev Temperature Expt.

Thermistor mounted to coil used to measure temperature during levitation



# Nonlinear Model

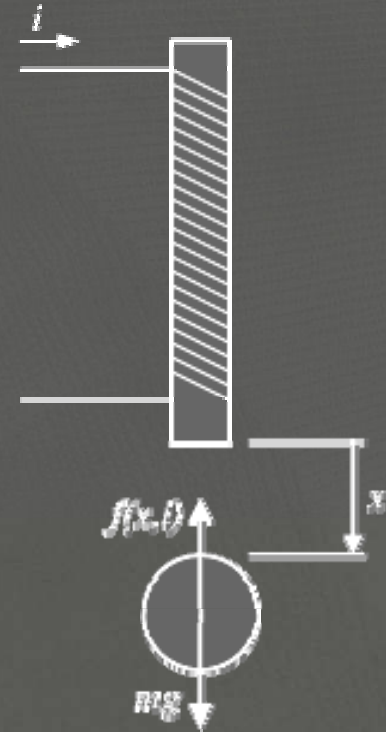
$$f(x, i, t) = -\frac{i(t)^2}{2} \frac{dL(x)}{dx}$$

$$L(x) = L_1 + \frac{L_o x_o}{x}$$

$$f(x, i, t) = \frac{L_o x_o}{2} \left( \frac{i(t)}{x(t)} \right)^2 = C \left( \frac{i(t)}{x(t)} \right)^2$$

$$m \frac{d^2 x(t)}{dt^2} = mg - f(x, i, t)$$

$$m \frac{d^2 x(t)}{dt^2} = mg - C \left( \frac{i(t)}{x(t)} \right)^2$$





# Linearized Model

$$f(x, i, t) = C \left( \frac{i_o}{x_o} \right)^2 + \left( \frac{2Ci_o}{x_o^2} \right) i(t) - \left( \frac{2Ci_o^2}{x_o^3} \right) x(t) + \dots$$

$$f(x, i, t) = f_o + f_1 + \dots$$

$$f_o(x, i) = C \left( \frac{i_o}{x_o} \right)^2 = mg$$

$$f_1(x, i, t) = \left( \frac{2Ci_o}{x_o^2} \right) i(t) - \left( \frac{2Ci_o^2}{x_o^3} \right) x(t)$$

$$m \frac{d^2 x(t)}{dt^2} = mg - f$$

$$m \frac{d^2 x(t)}{dt^2} = mg - f_o - \left( \frac{2Ci_o}{x_o^2} \right) i(t) + \left( \frac{2Ci_o^2}{x_o^3} \right) x(t)$$

# Linearized Model (cont'd)

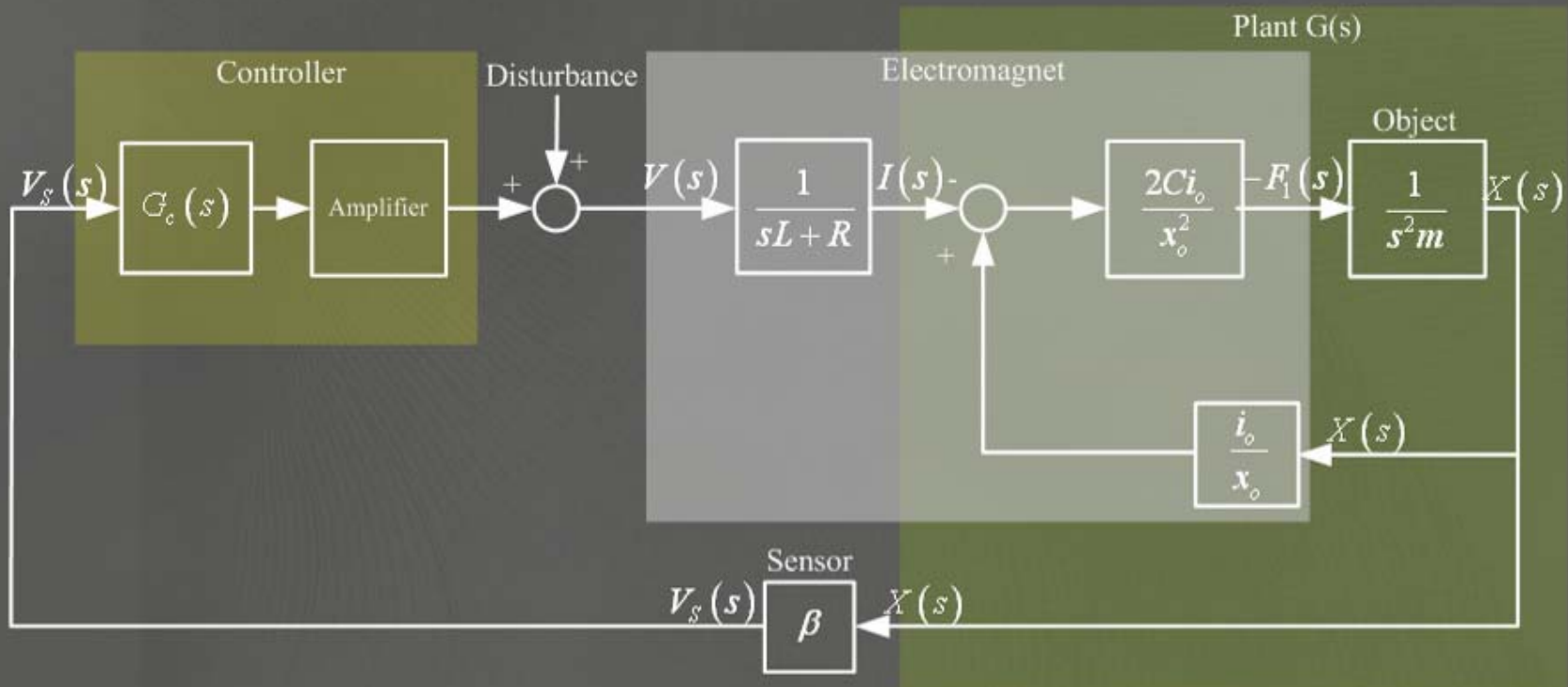
$$f_o = C \left( \frac{i_o}{x_o} \right)^2 = mg$$

$$m \frac{d^2 x(t)}{dt^2} = - \left( \frac{2Ci_o}{x_o^2} \right) i(t) + \left( \frac{2Ci_o^2}{x_o^3} \right) x(t)$$

$$ms^2 X(s) = - \left( \frac{2Ci_o}{x_o^2} \right) I(s) + \left( \frac{2Ci_o^2}{x_o^3} \right) X(s)$$

$$G(s) = \frac{X(s)}{I(s)} = \frac{- \left( \frac{2Ci_o}{x_o^2} \right)}{ms^2 - \left( \frac{2Ci_o^2}{x_o^3} \right)} = \frac{- \left( \frac{2Ci_o}{mx_o^2} \right)}{s^2 - \left( \frac{2Ci_o^2}{mx_o^3} \right)}$$

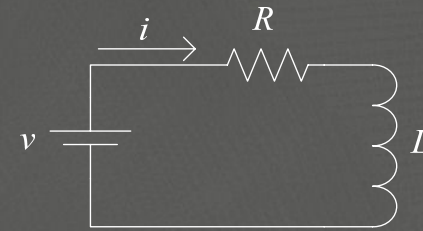
# Block Diagram



$$G(s) = \frac{X(s)}{I(s)} = \frac{-\left(\frac{2Ci_o}{x_o^2}\right)}{s^2m - \left(\frac{2Ci_o^2}{x_o^3}\right)} = \frac{-\left(\frac{2Ci_o}{mx_o^2}\right)}{s^2 - \left(\frac{2Ci_o^2}{mx_o^3}\right)}$$

# Alternate Plant Transfer Function

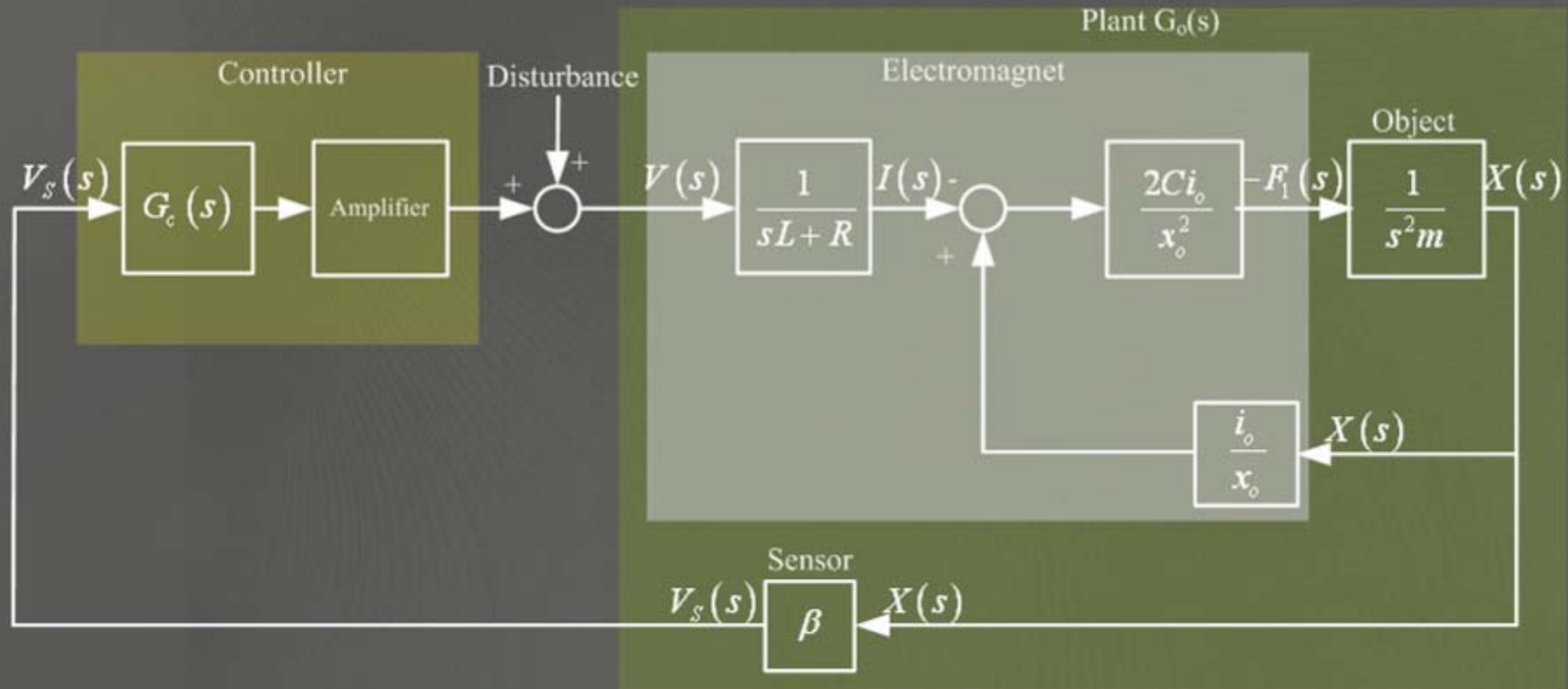
- o The electromagnet represented as a series combination of a resistor and inductor
- o Consider sensor output as the system output
- o Consider voltage to the electromagnet as the system input



$$v(x, i, t) = Ri(t) + L(x) \frac{di(t)}{dt}$$

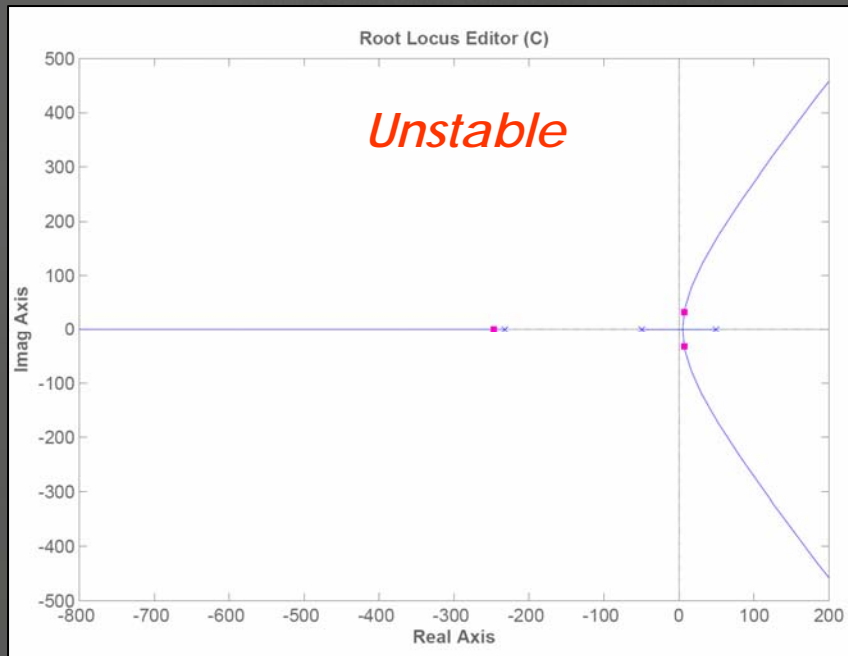
$$G_o(s) = \frac{V_s(s)}{V(s)} = \frac{\left( \frac{-2\beta C i_o}{mLx_o^2} \right)}{\left( s + \frac{R}{L} \right) \left( s^2 - \frac{2C i_o^2}{mx_o^3} \right)}$$

# Block Diagram II



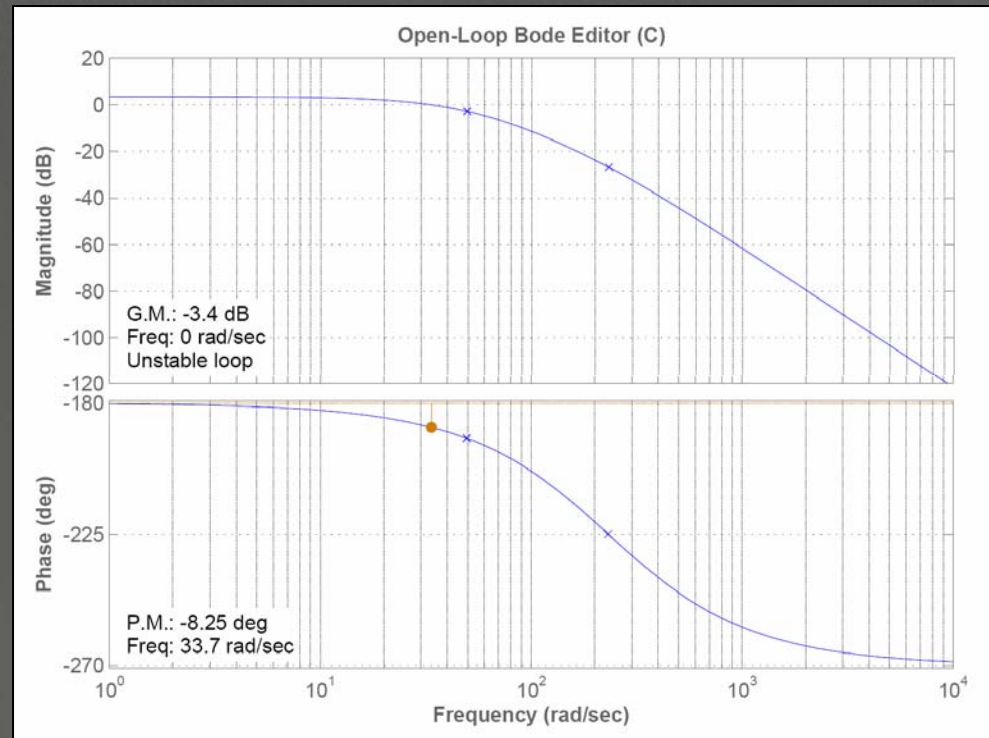
$$G_o(s) = \frac{V_s(s)}{V(s)} = \frac{\left( \frac{-2\beta C i_o}{m L x_o^2} \right)}{\left( s + \frac{R}{L} \right) \left( s^2 - \frac{2C i_o^2}{m x_o^3} \right)}$$

# Open-loop Analysis



The root-locus has a pole in the right-half plane; the system is **unstable**

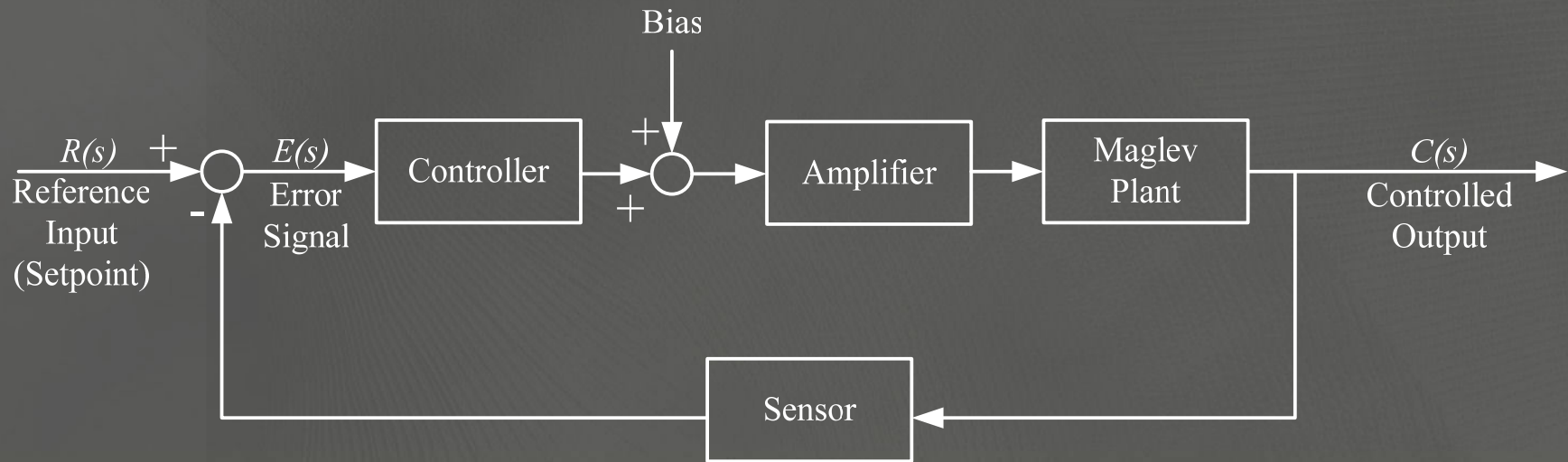
$$G_o(s) = \frac{824000}{(s + 232)(s + 49.5)(s - 49.5)}$$



# Outline of Presentation

- Objectives
- Background
- Maglev Testbed
- **Control Strategies**
- Simulation Studies
- Experimental Studies
- Summary
- Future Work

# Control Block Diagram



- o Error signal calculated as difference of setpoint and sensor output
- o Error signal is input to controller
- o Bias is added to controller output, amplified, and input to “maglev plant”
- o Controlled output is actual output



$$G_{Lead/Lag}(s) = K \frac{s+z}{s+p} \approx K(s+z) \quad (p \rightarrow \infty)$$

# Control Strategies

## o Proportional Control

- o Controller output is proportional to error

$$G_P(s) = K_P$$

## o Lead-Lag Control

- o Adds a zero and a pole to the open-loop system

$$G_{Lead/Lag}(s) = K \frac{s+z}{s+p} \quad \begin{array}{l} \text{Lead for } 0 < z < p \\ \text{Lag for } 0 < p < z \end{array}$$

# Control Strategies

- PD Control

- Adds derivative term

$$G_{PD}(s) = K_P + K_D s$$

- PID Control

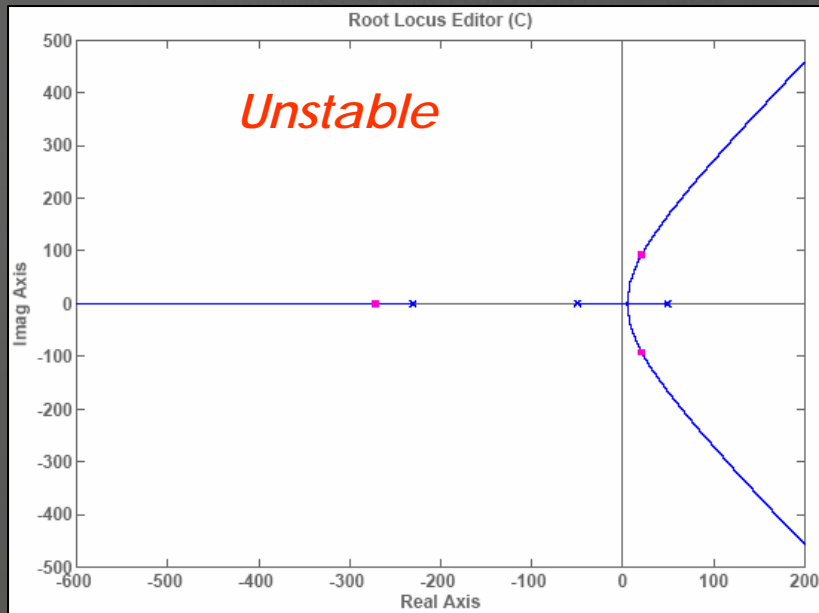
- Integral term makes steady-state error zero

$$G_{PID}(s) = K_P + K_D s + \frac{K_I}{s}$$

# Outline of Presentation

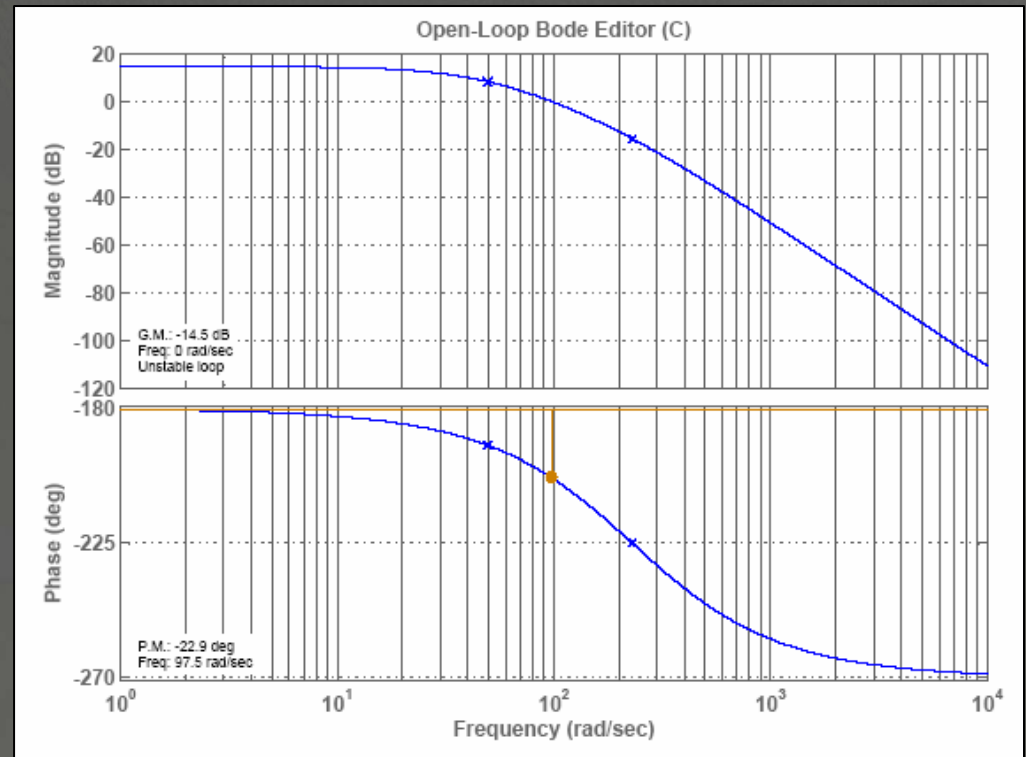
- Objectives
- Background
- Maglev Testbed
- Control Strategies
- **Simulation Studies**
- Experimental Studies
- Summary
- Future Work

# Proportional Controller

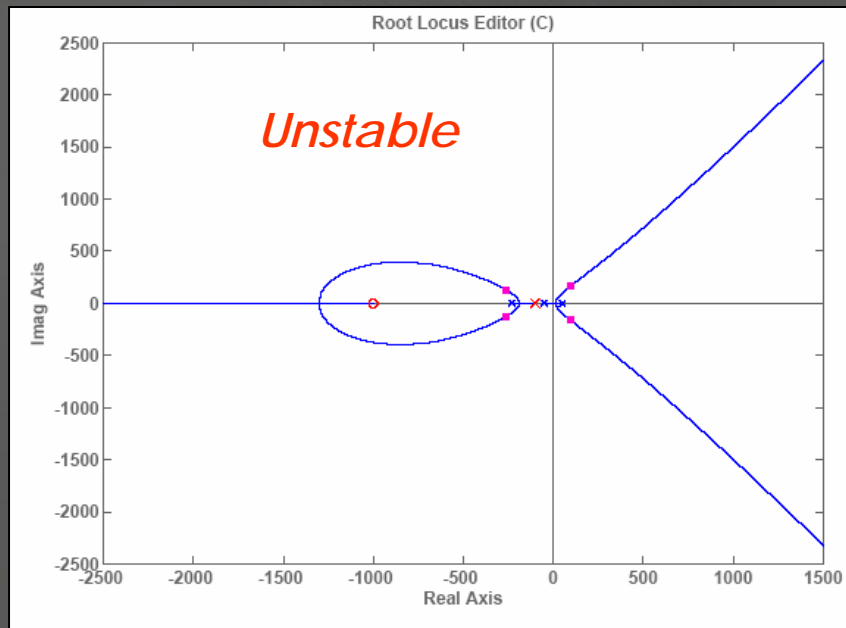


Proportional Controller by itself cannot stabilize open-loop unstable system

Proportional Controller acts as a gain; cannot move poles into left-half plane

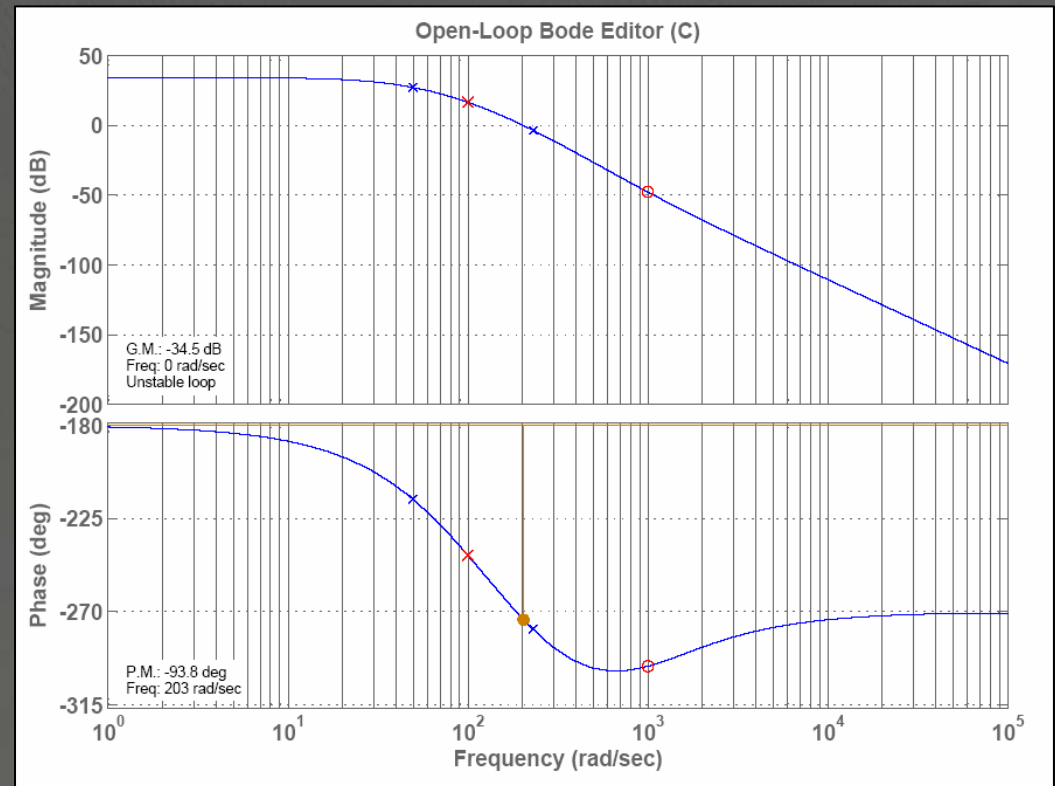


# Lag Compensator

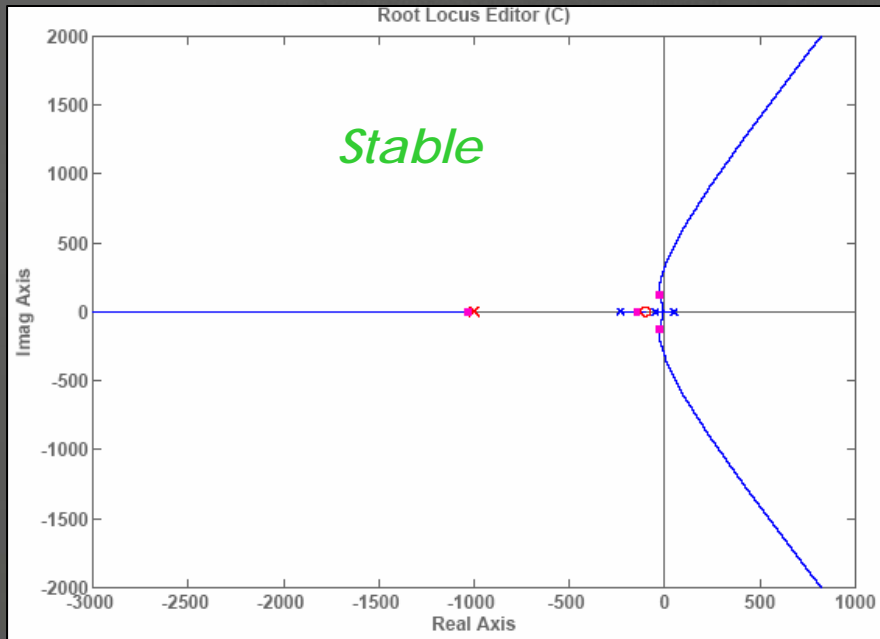


Lag Compensator cannot move poles into left-half plane

Lag Compensator by itself cannot stabilize an open-loop unstable system



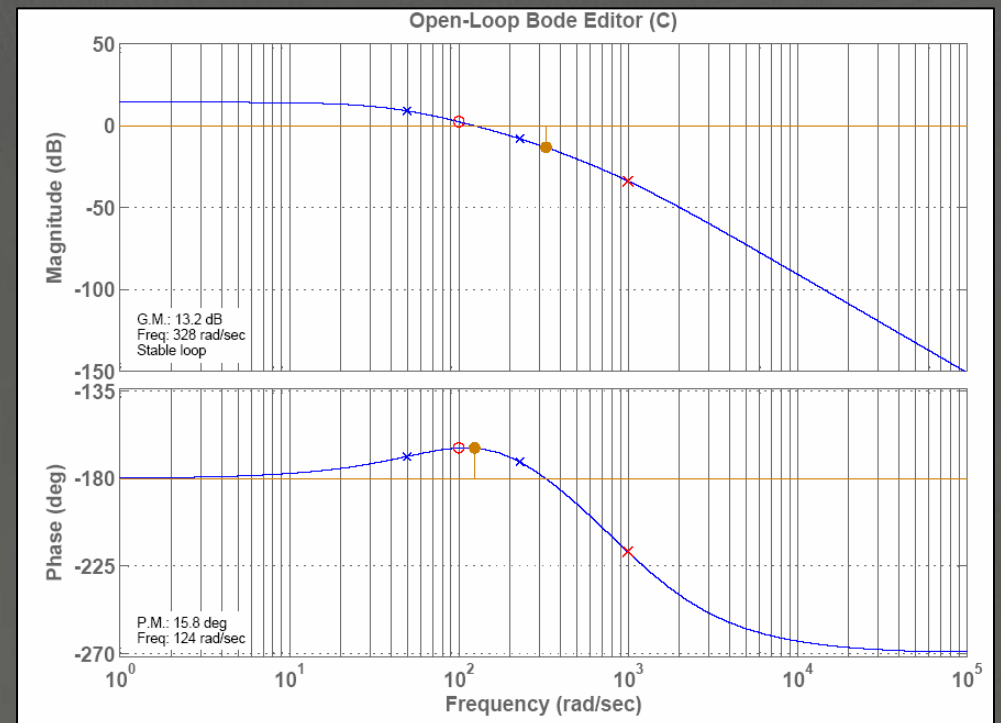
# Lead Compensator



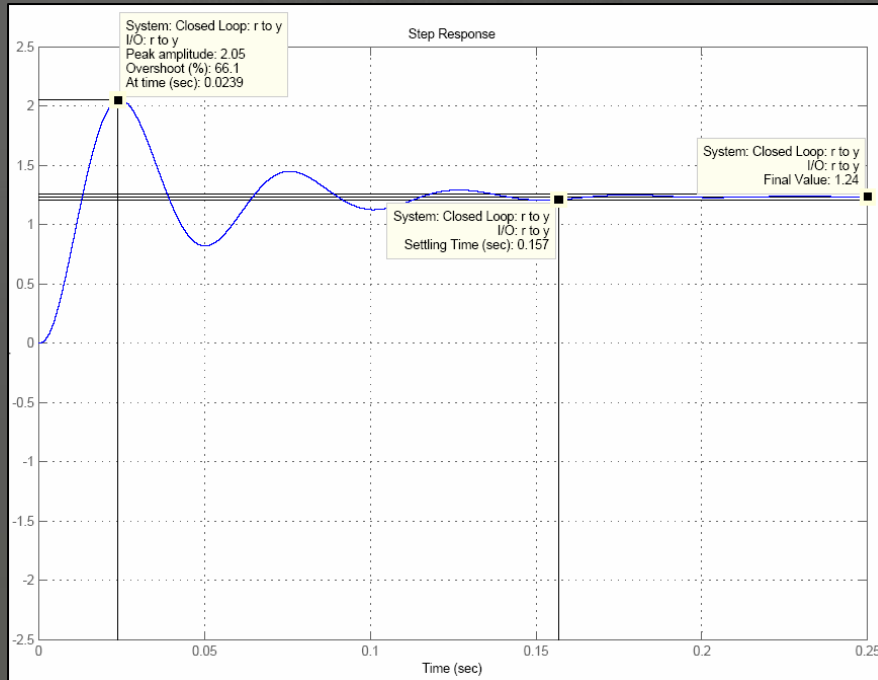
Lead Compensator  
can stabilize open-  
loop unstable system

Gain Margin: 13.2 dB  
Phase Margin: 15.8 deg

Lead Compensator  
moves poles into  
left-half plane



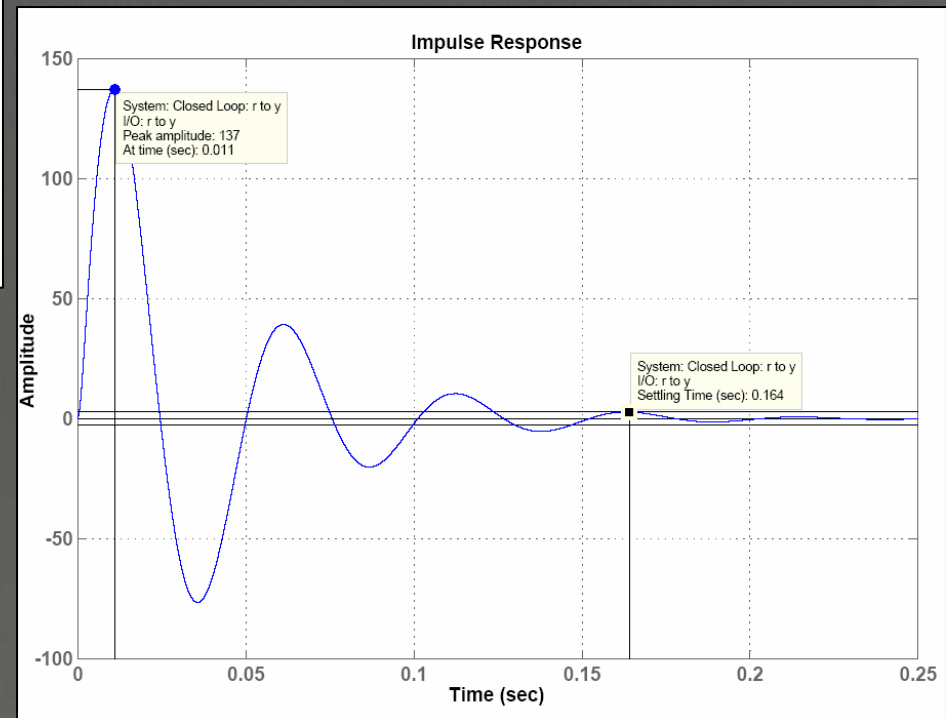
# Lead Compensator Response



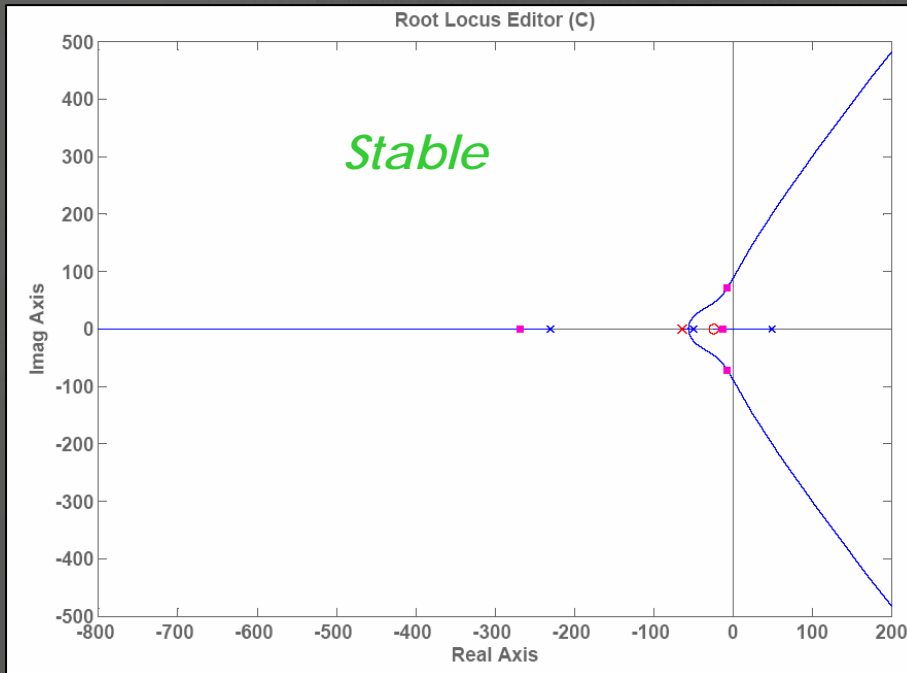
Step Response  
Settling Time: 0.157s  
Steady-State Error: 0.24

(Zero at -100 rad/s, Pole at -1000 rad/s)

Impulse Response  
Settling Time: 0.164s



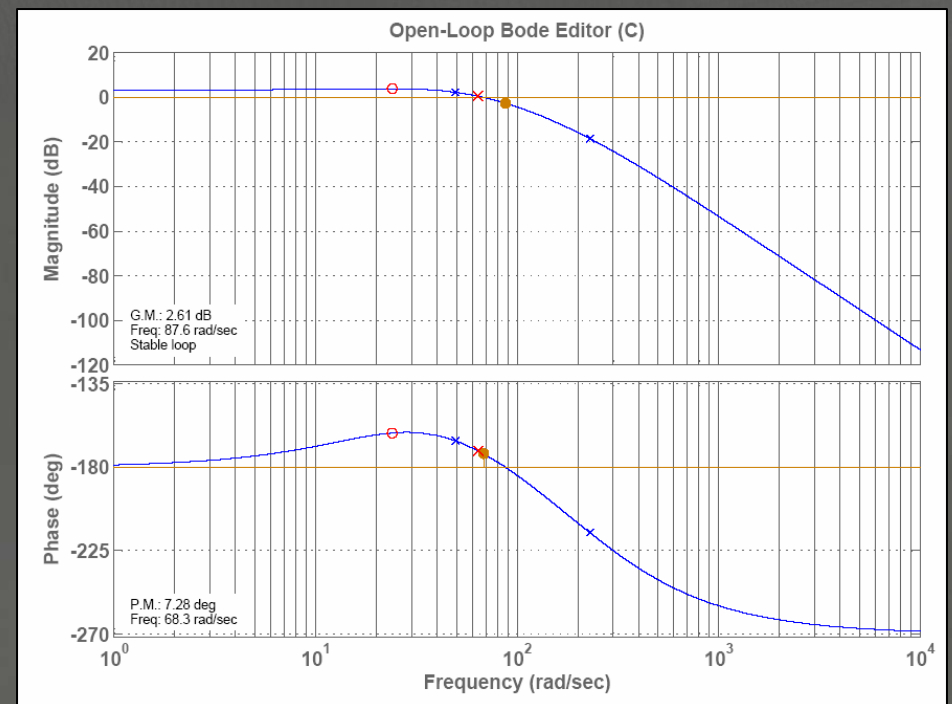
# PD Controller



PD Controller  
moves poles into  
left-half plane

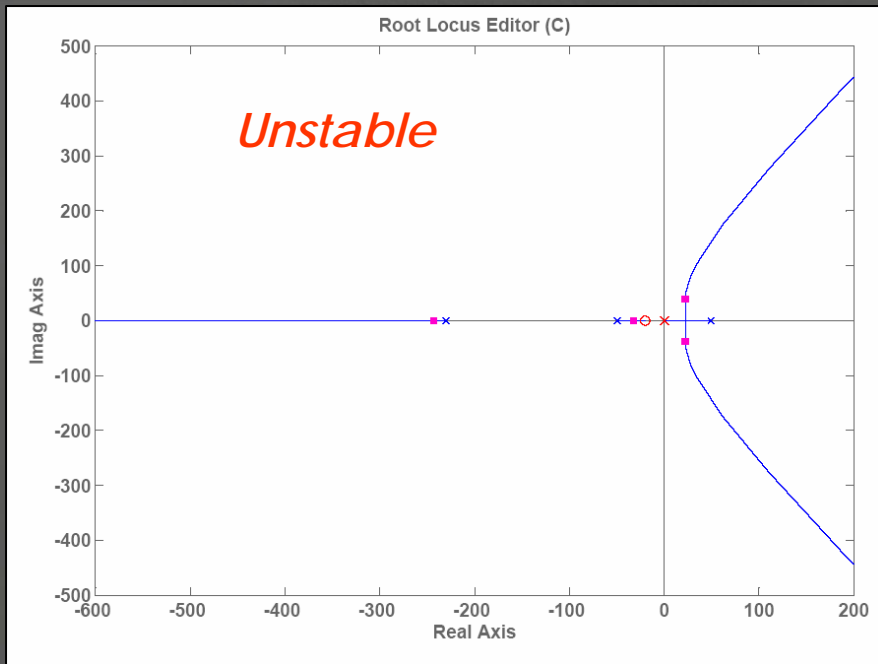
PD Controller can  
stabilize open-loop  
unstable system

Gain Margin: 2.61 dB  
Phase Margin: 7.28 deg



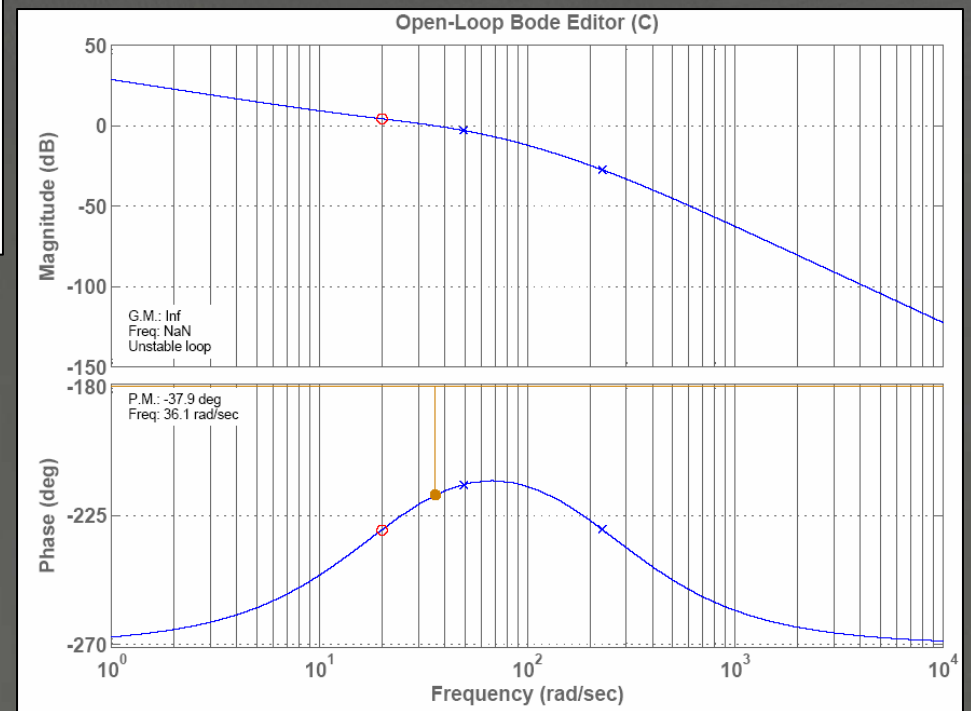


# PI Controller

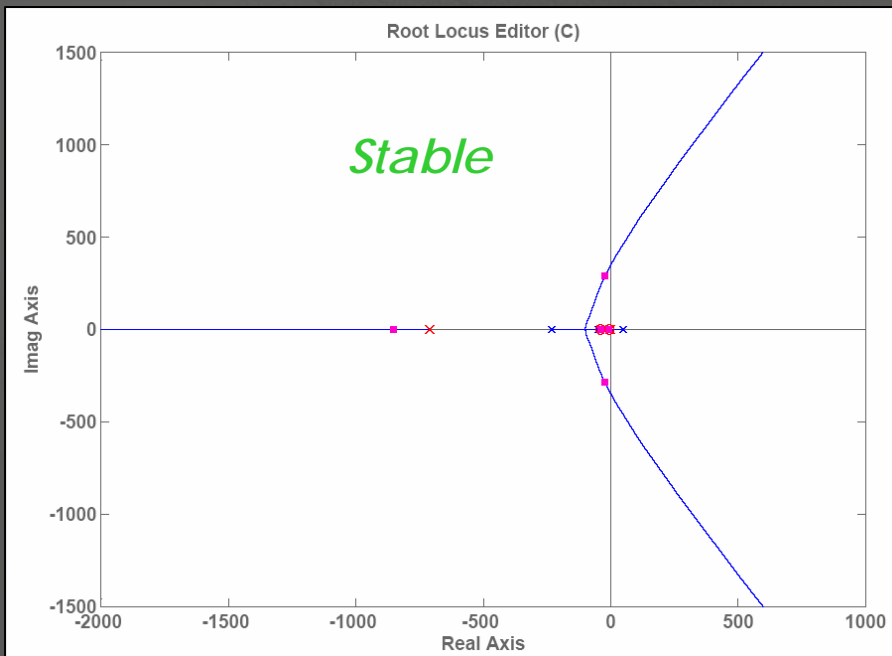


PI Controller cannot move poles into left-half plane

PI Controller cannot stabilize open-loop unstable system



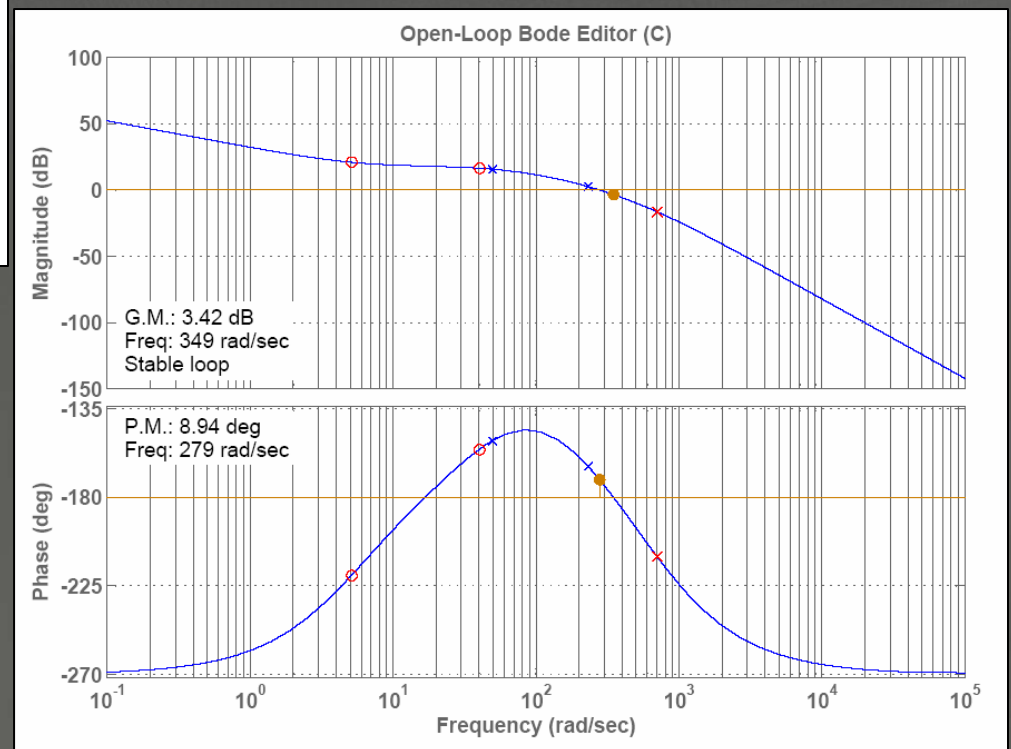
# PID Controller



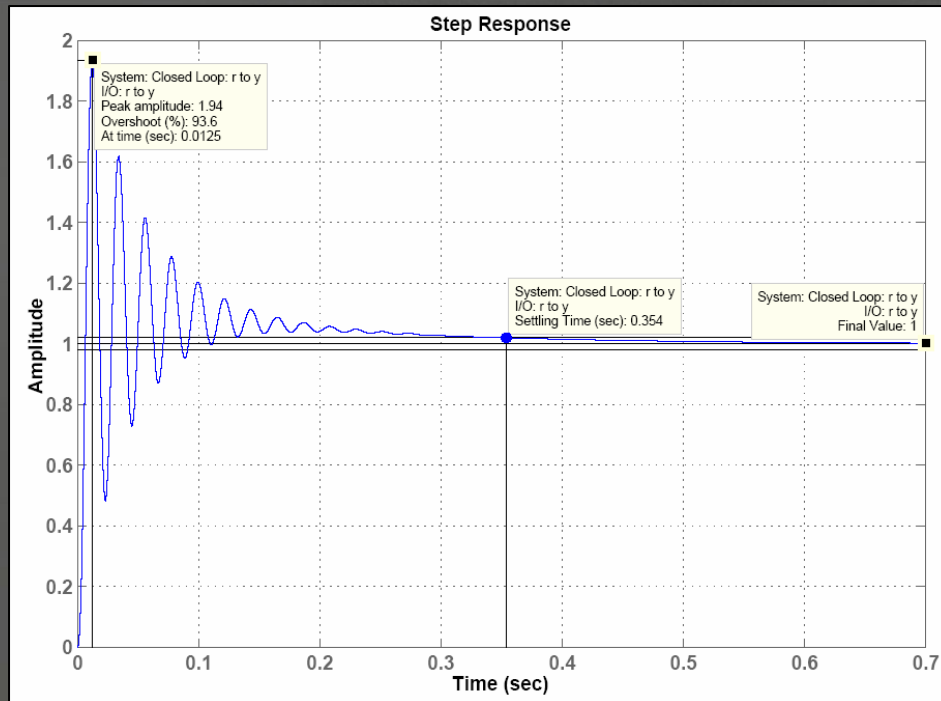
PID Controller  
moves poles into  
left-half plane

PID Controller can  
stabilize open-loop  
unstable system

Gain Margin: 3.42 dB  
Phase Margin: 8.94 deg



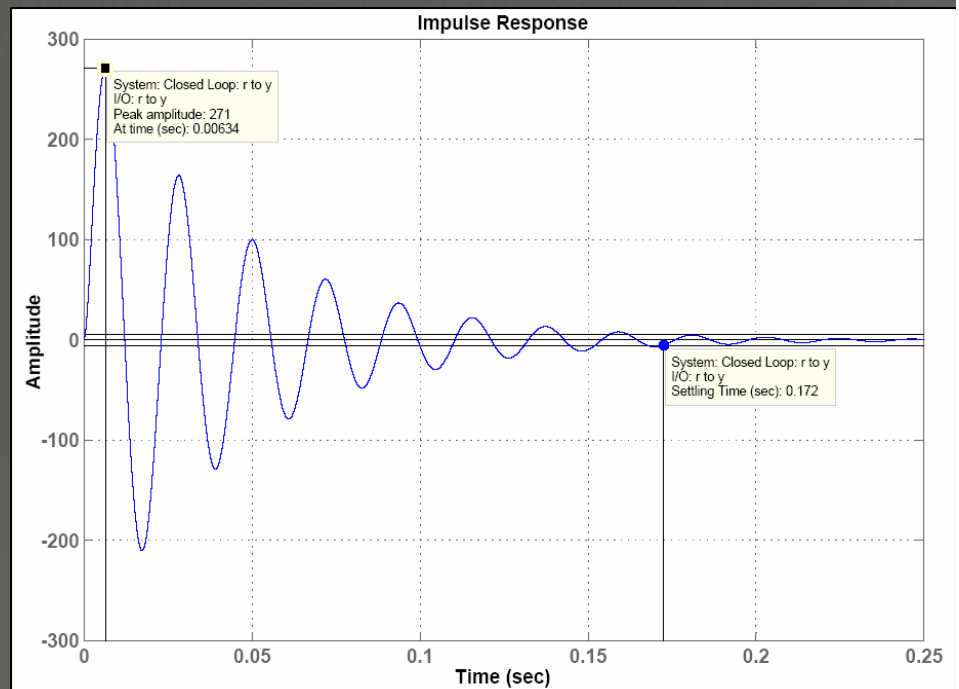
# PID Controller Response



Step Response  
Settling Time: 0.354s  
Steady-State Error: 0

$$(K_P=0.68, K_I=3.12, K_D=0.0141)$$

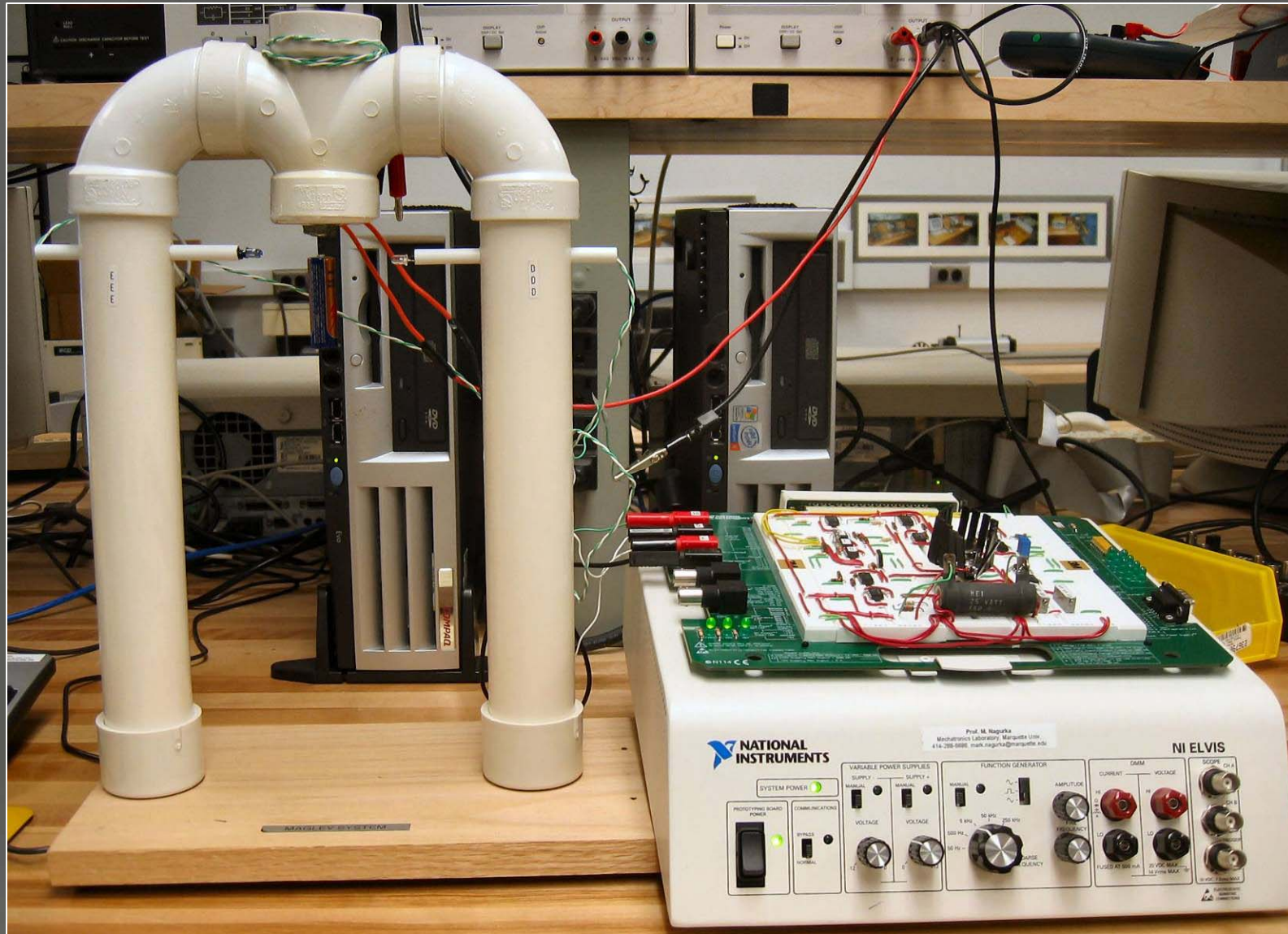
Impulse Response  
Settling Time: 0.172s



# Outline of Presentation

- Objectives
- Background
- Maglev Testbed
- Control Strategies
- Simulation Studies
- **Experimental Studies**
- Summary
- Future Work

# Maglev Experimental Station



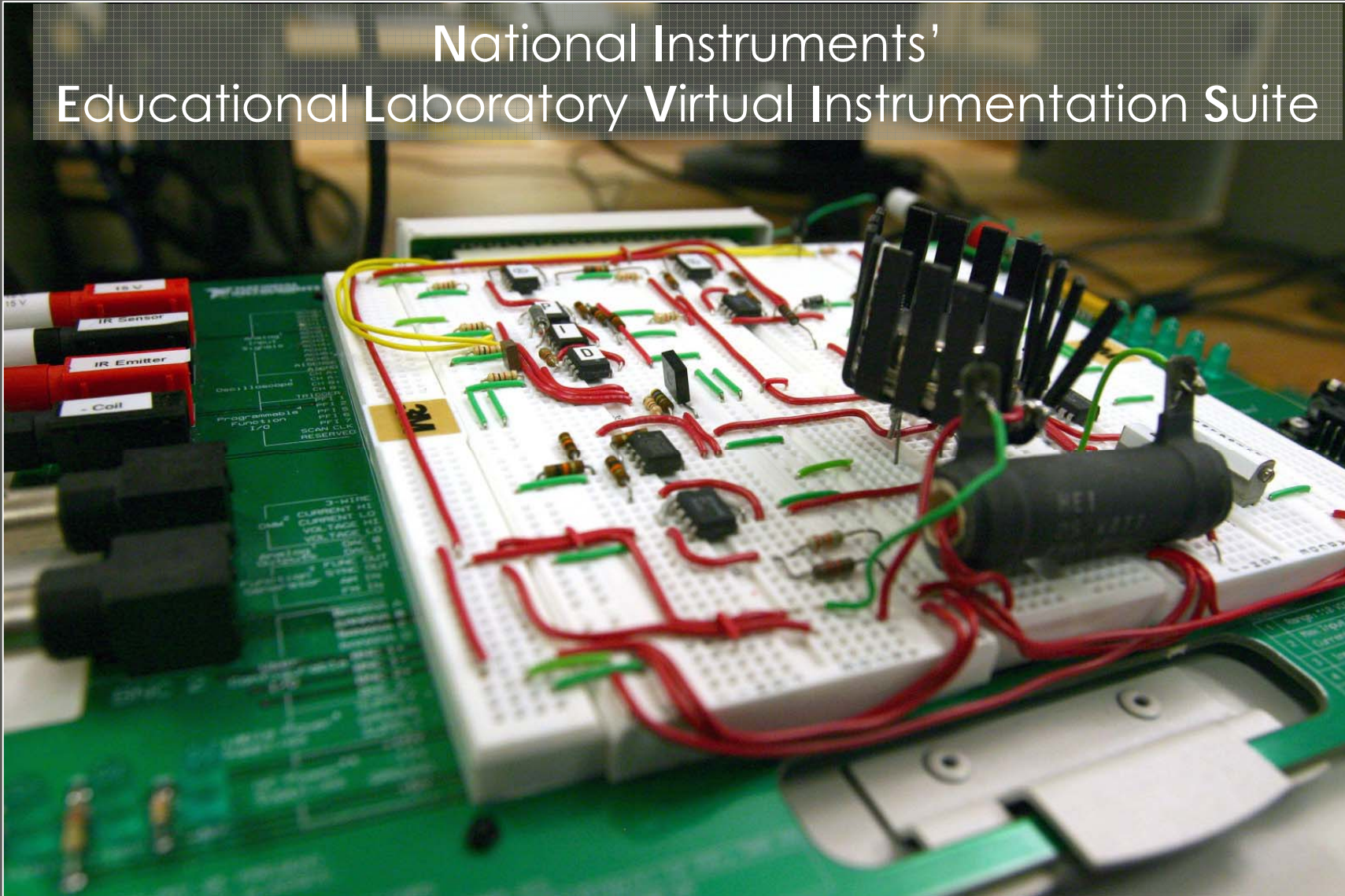
# NI-ELVIS

National Instruments'  
Educational Laboratory Virtual Instrumentation Suite

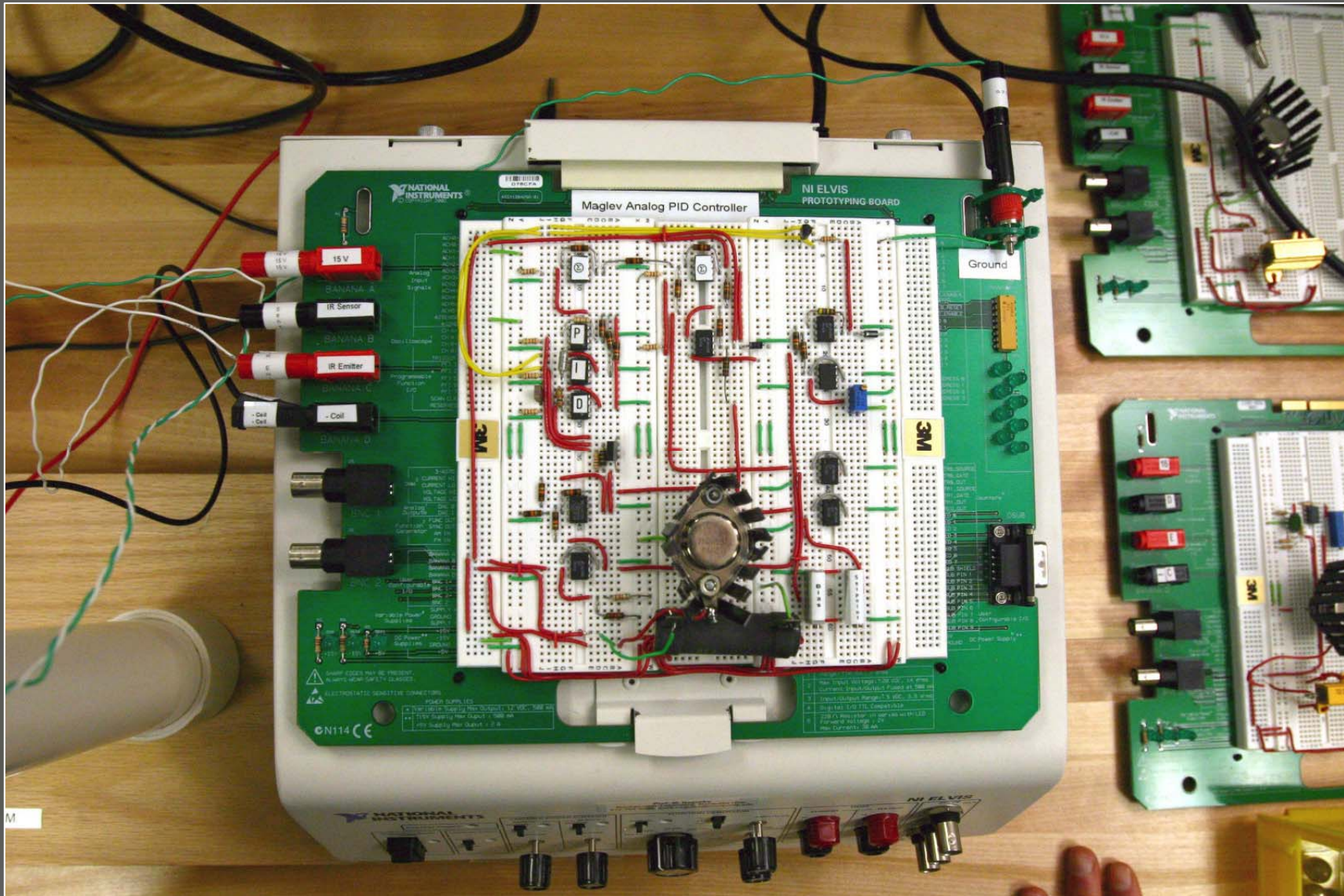


# NI-ELVIS Prototyping Board

National Instruments'  
Educational Laboratory Virtual Instrumentation Suite

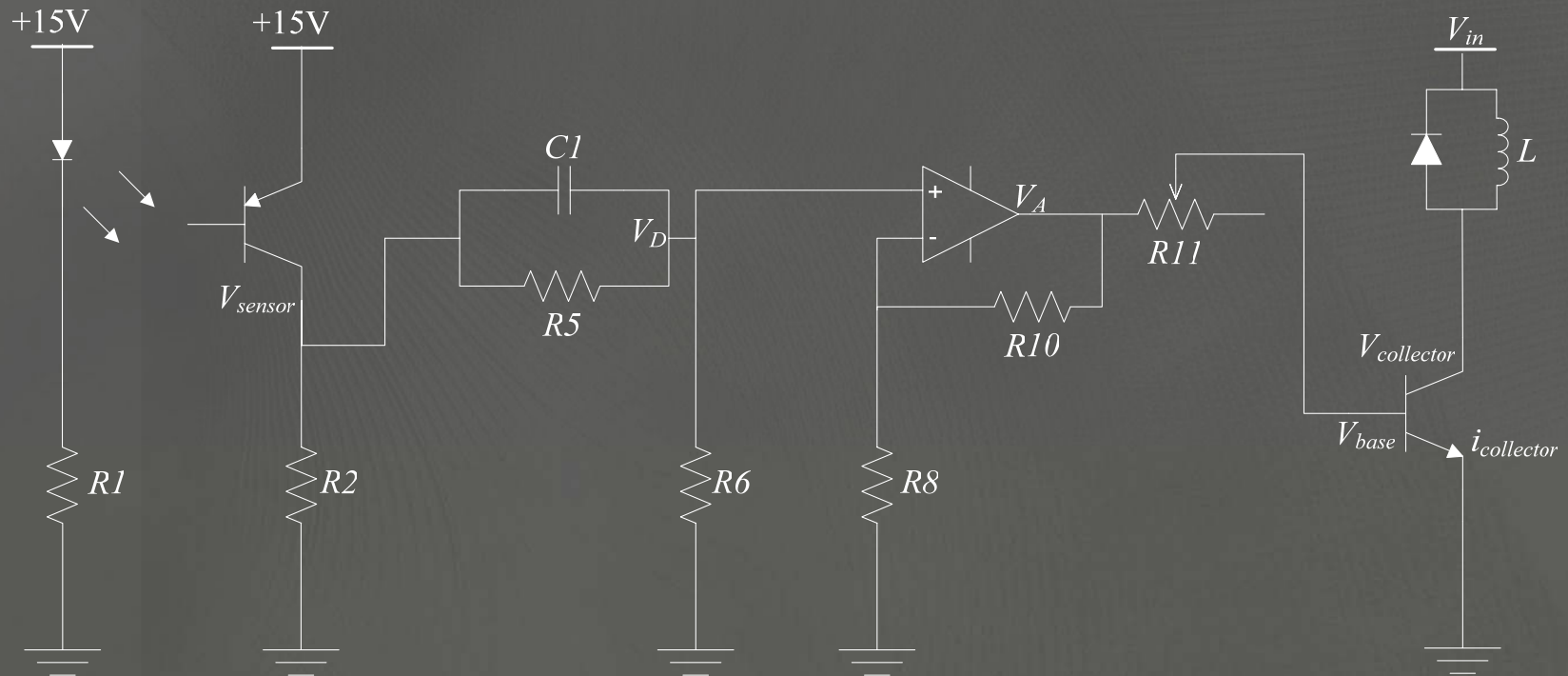


# Analog Control with NI-ELVIS



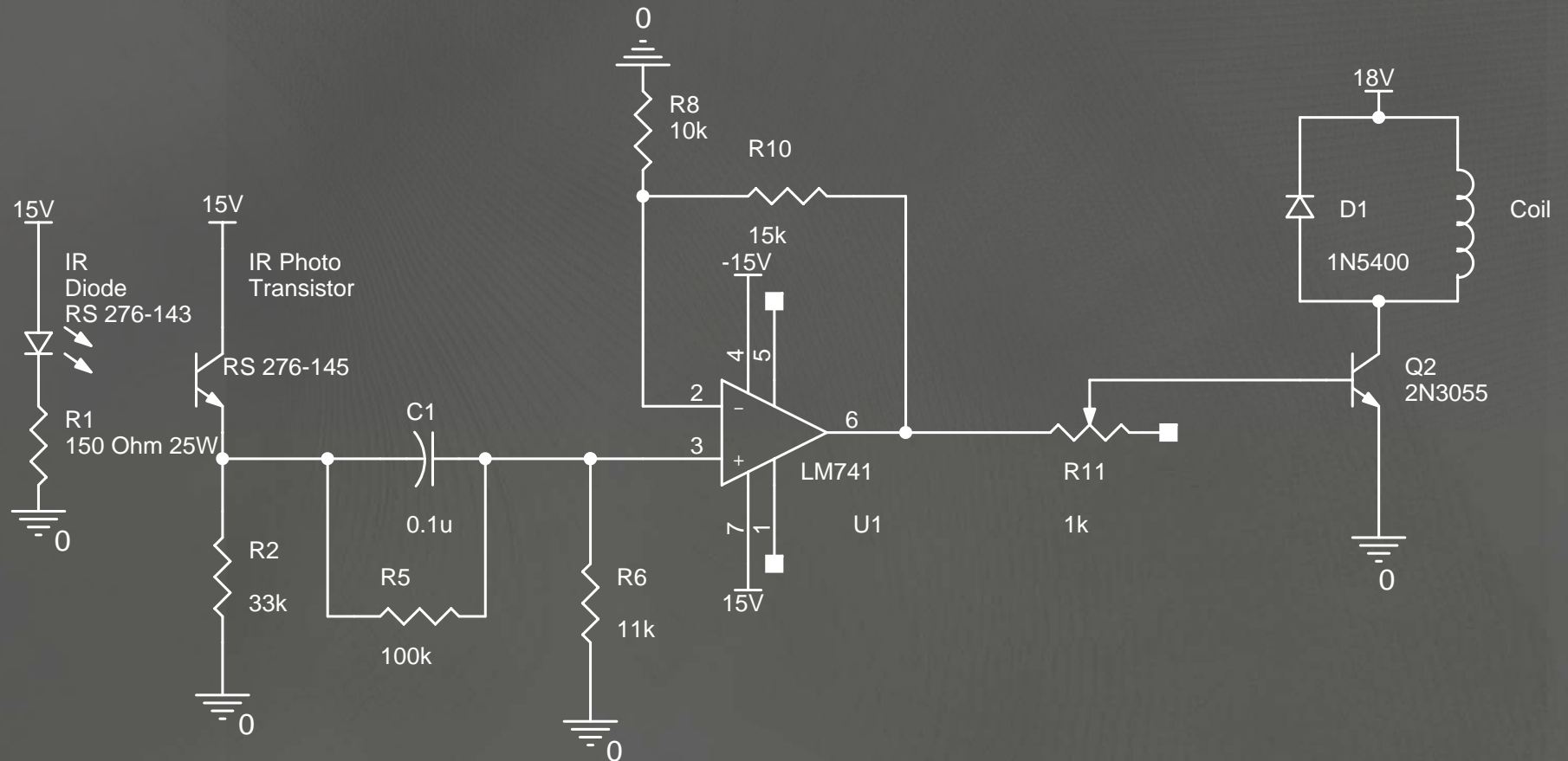


# Lead Compensator Analog Circuit

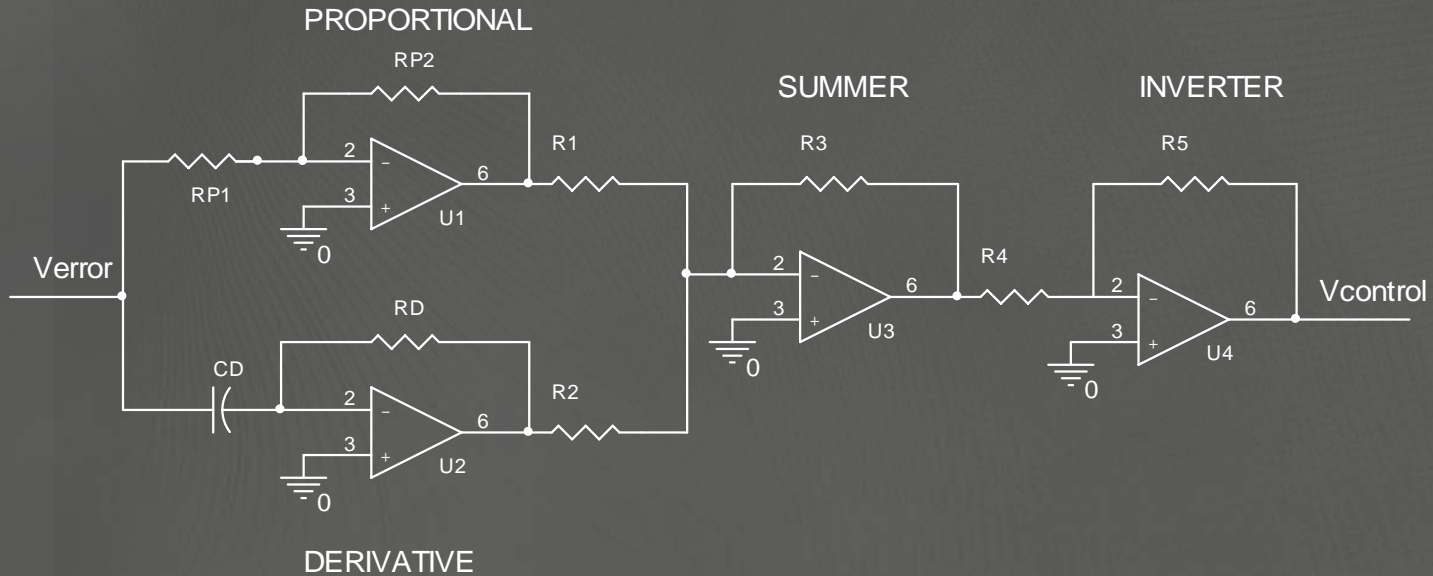


$$G(s) = \frac{V_D}{V_{sensor}} = \frac{s + \frac{1}{R_5 C_1}}{s + \frac{R_5 + R_6}{R_5 R_6 C_1}}$$

# Lead Compensator Analog Implementation

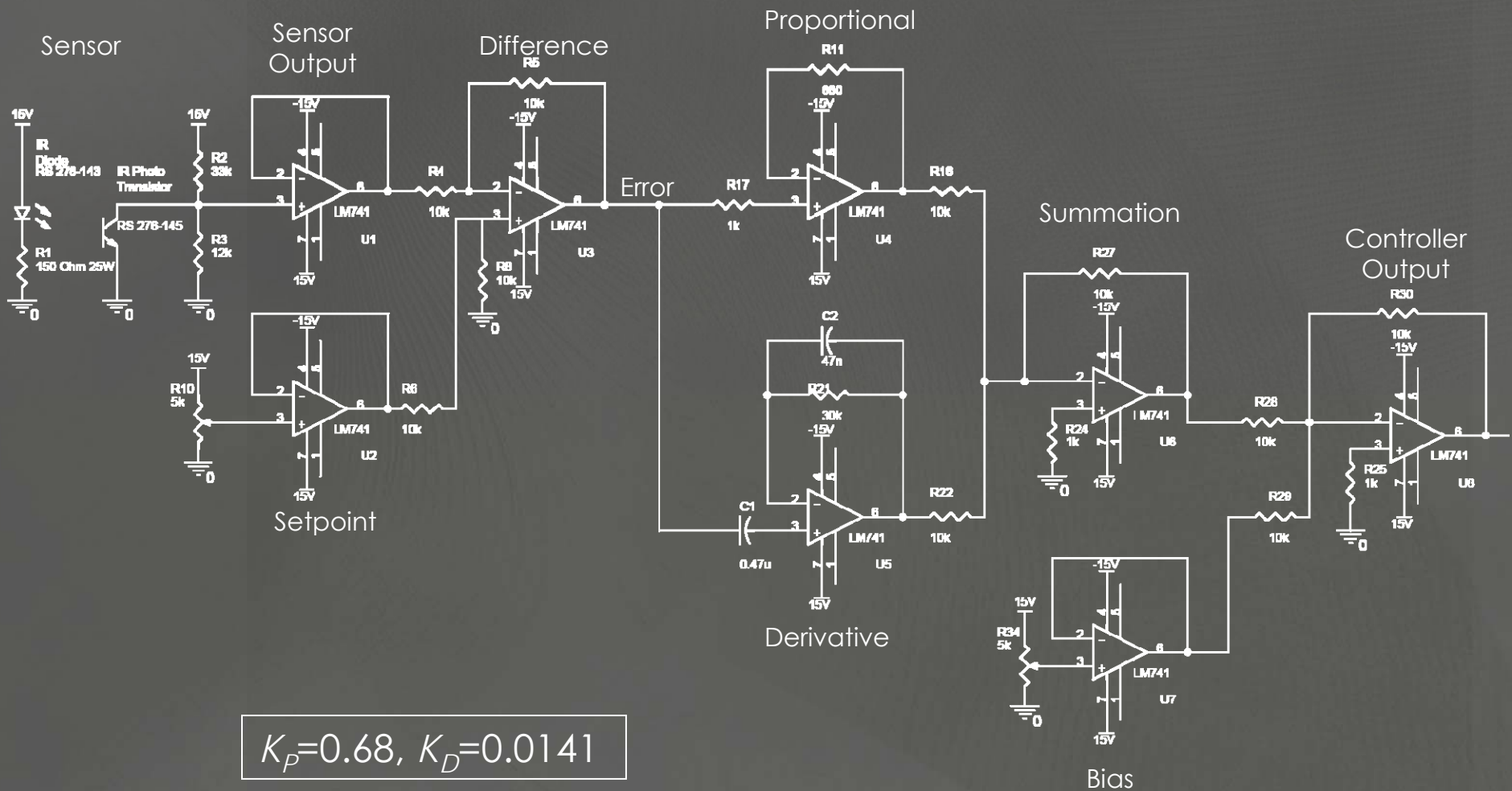


# PD Analog Circuit



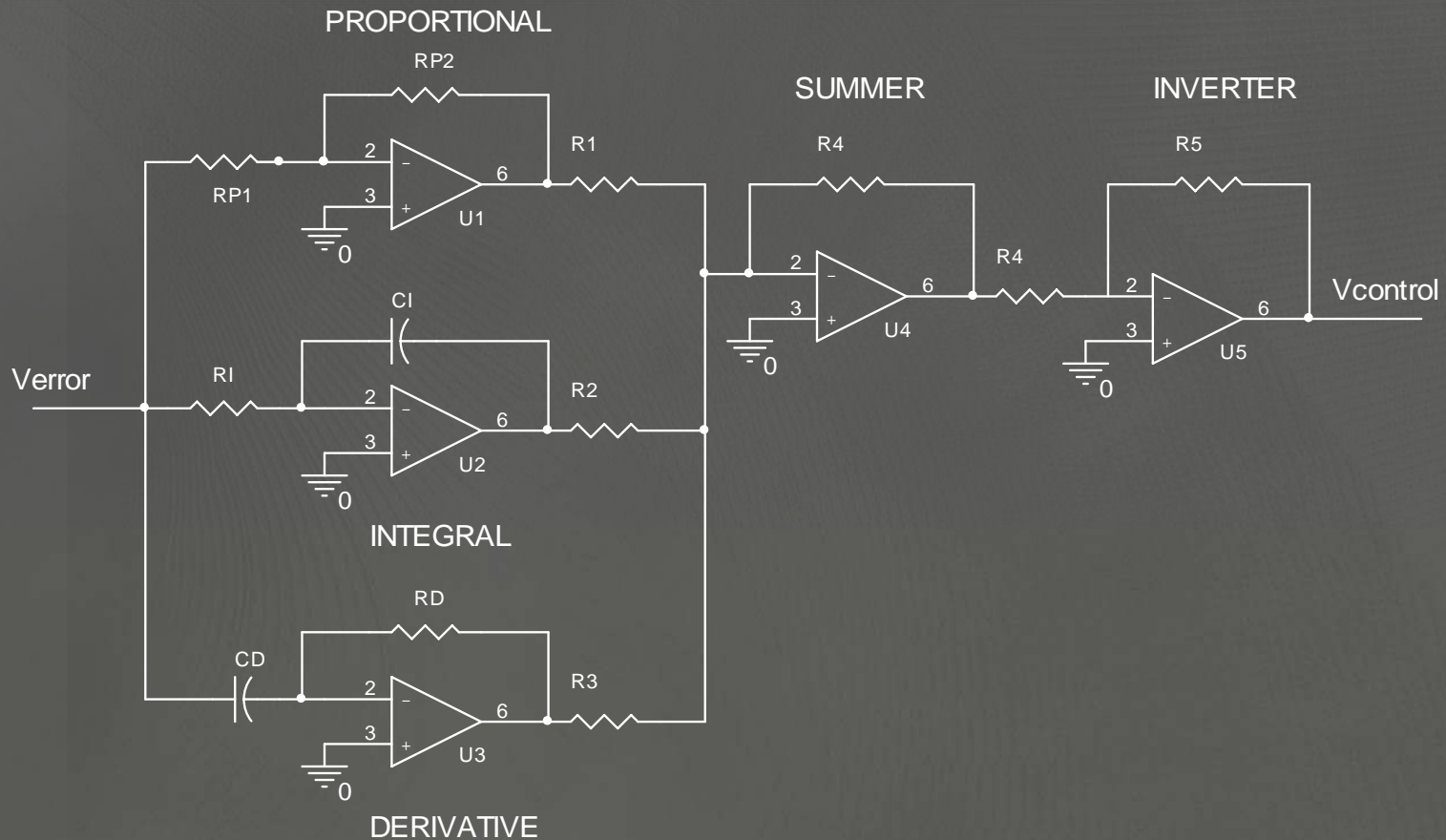
$$G(s) = \frac{V_{control}(s)}{V_{error}(s)} = \underbrace{\left(\frac{RP2}{RP1}\right)}_{K_P} + \underbrace{(RD \times CD)}_{K_D} s$$

# PD Analog Circuit Implementation



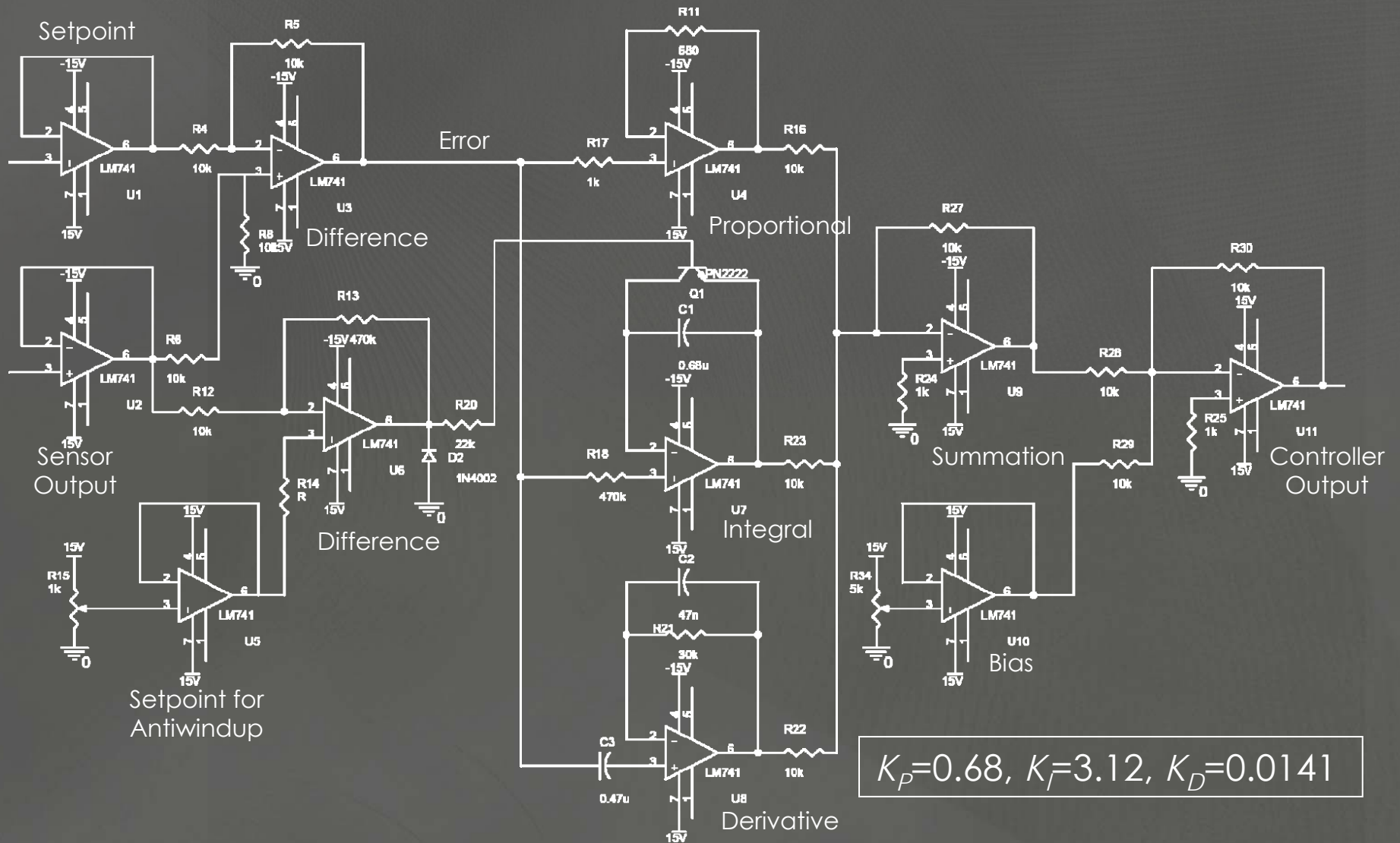
$$K_P=0.68, K_D=0.0141$$

# PID Analog Circuit

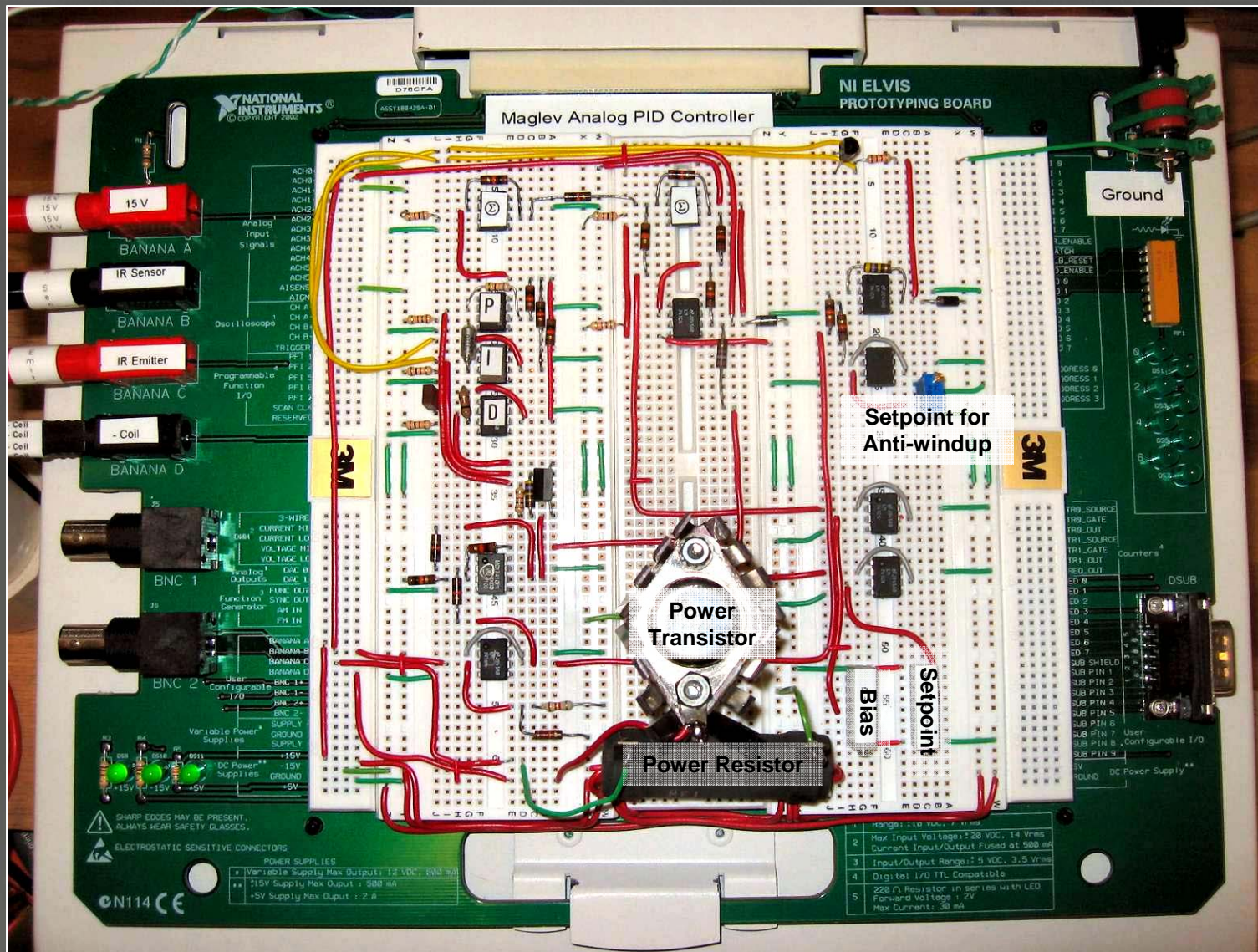


$$G(s) = \frac{V_{control}(s)}{V_{error}(s)} = \underbrace{\left(\frac{RP2}{RP1}\right)}_{K_P} + \underbrace{\left(\frac{1}{RI \times CI}\right)}_{K_I} \frac{1}{s} + \underbrace{(RD \times CD)}_{K_D} s$$

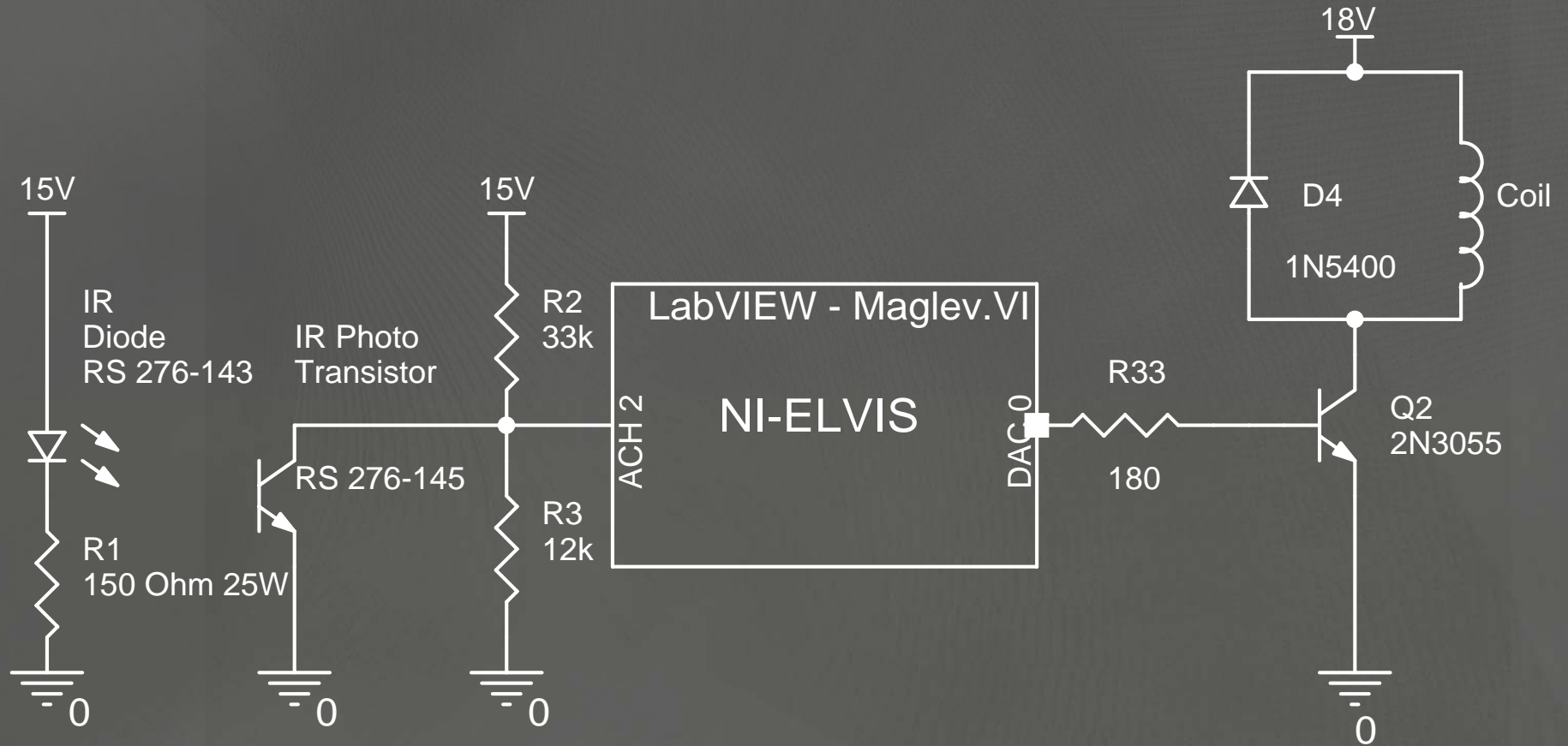
# PID Analog Circuit Implementation



# Analog Implementation

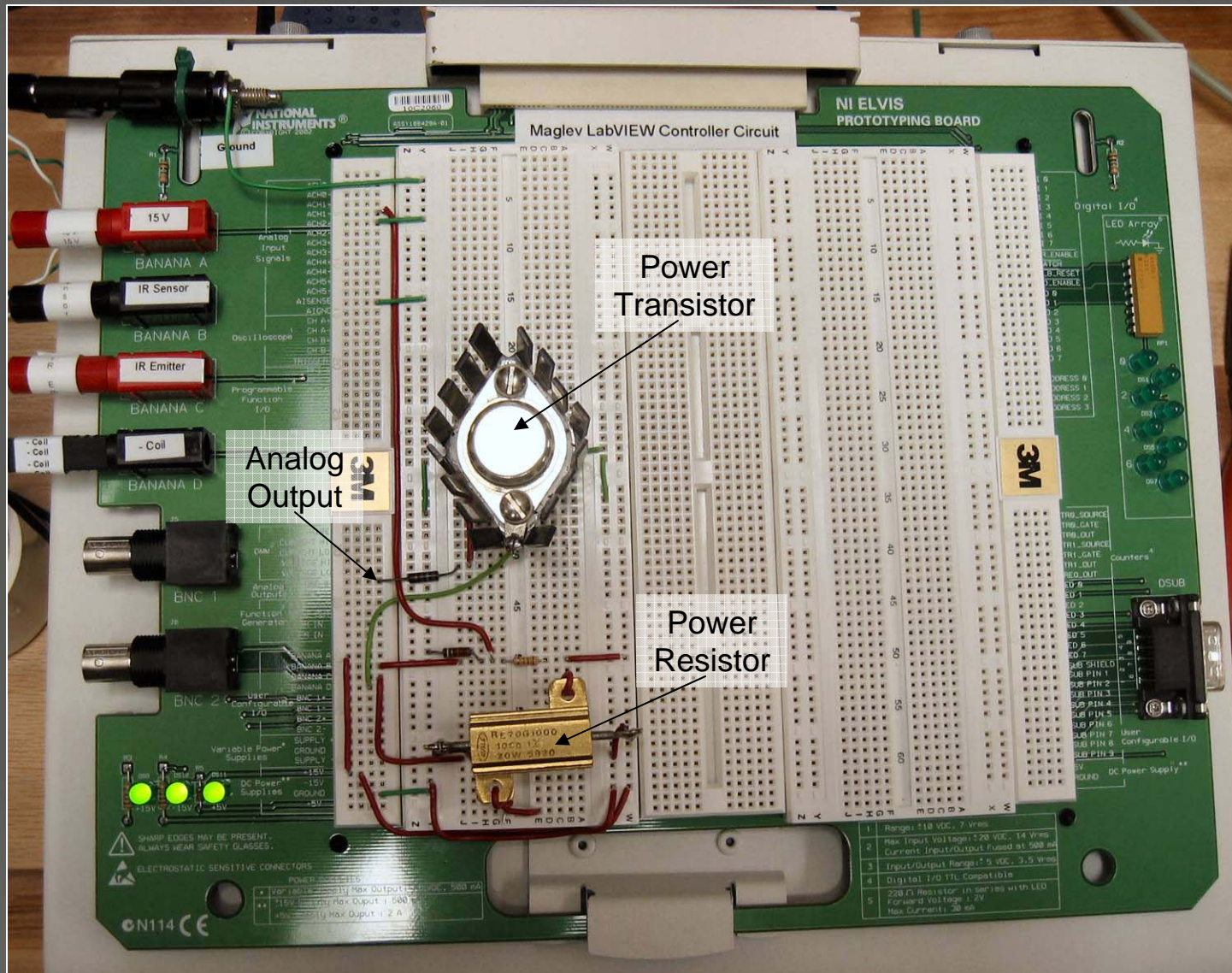


# LabVIEW-based Control

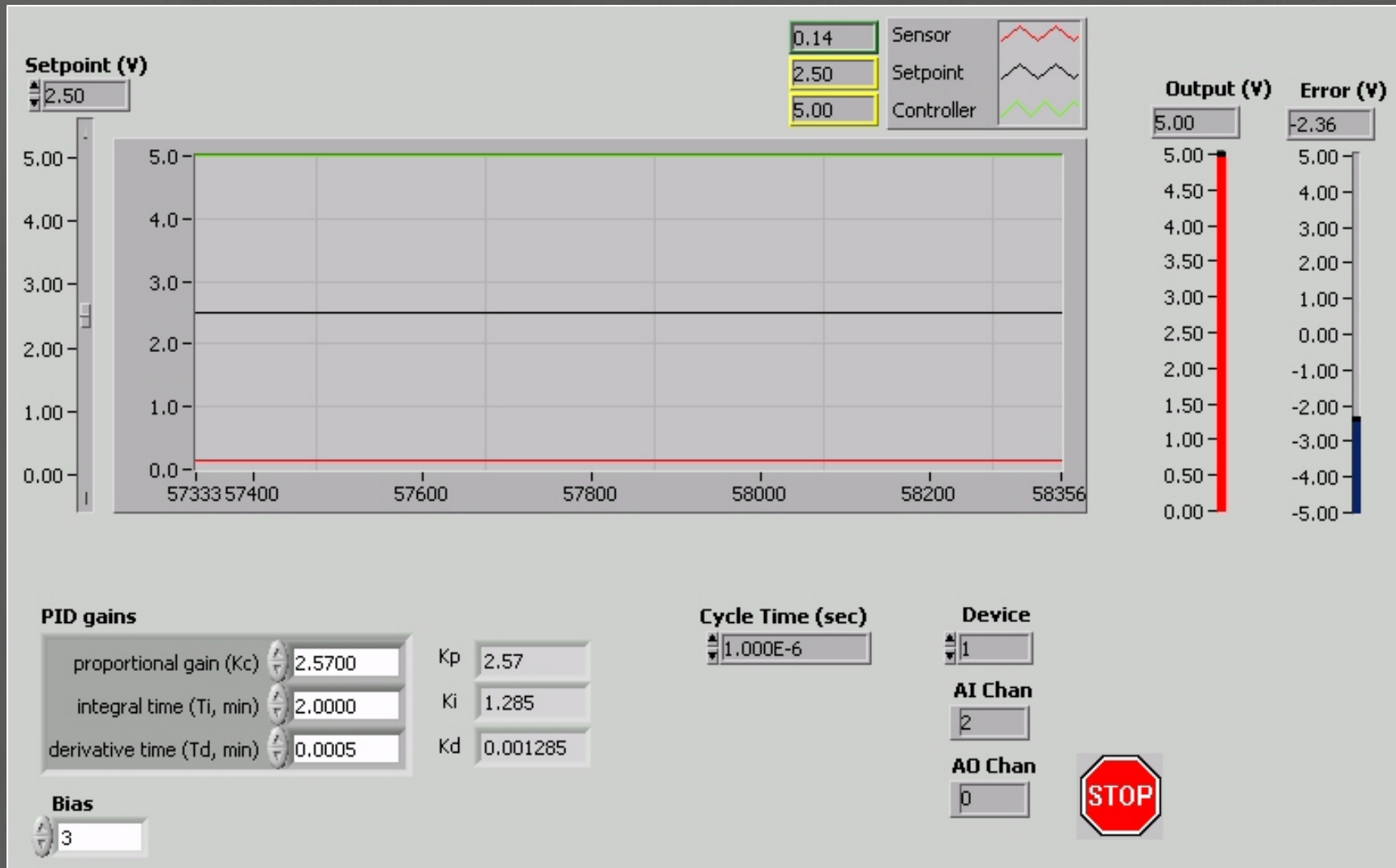




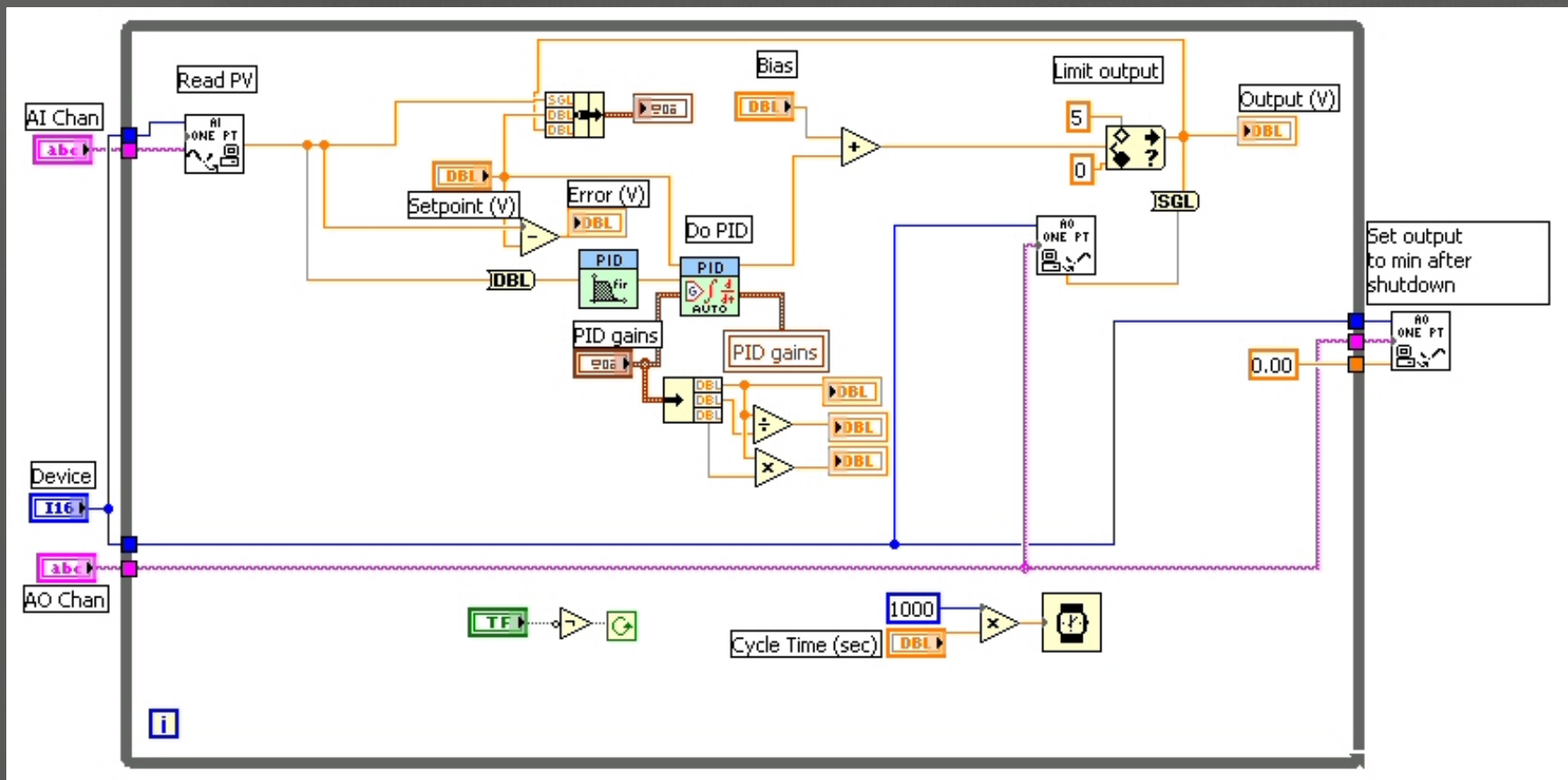
# LabVIEW-based Control



# LabVIEW VI Front Panel



# LabVIEW VI Program



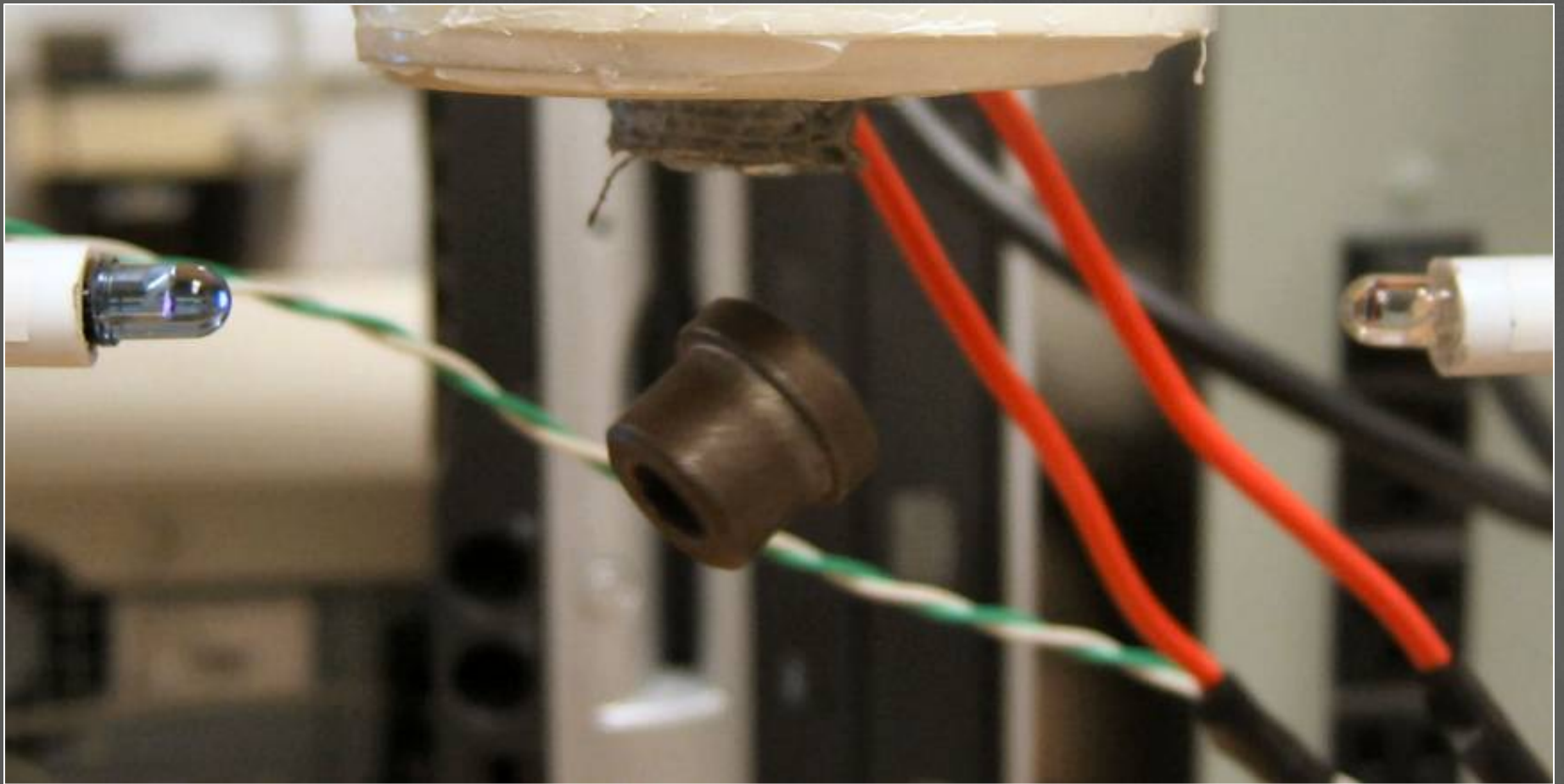
# Levitation of AA Battery



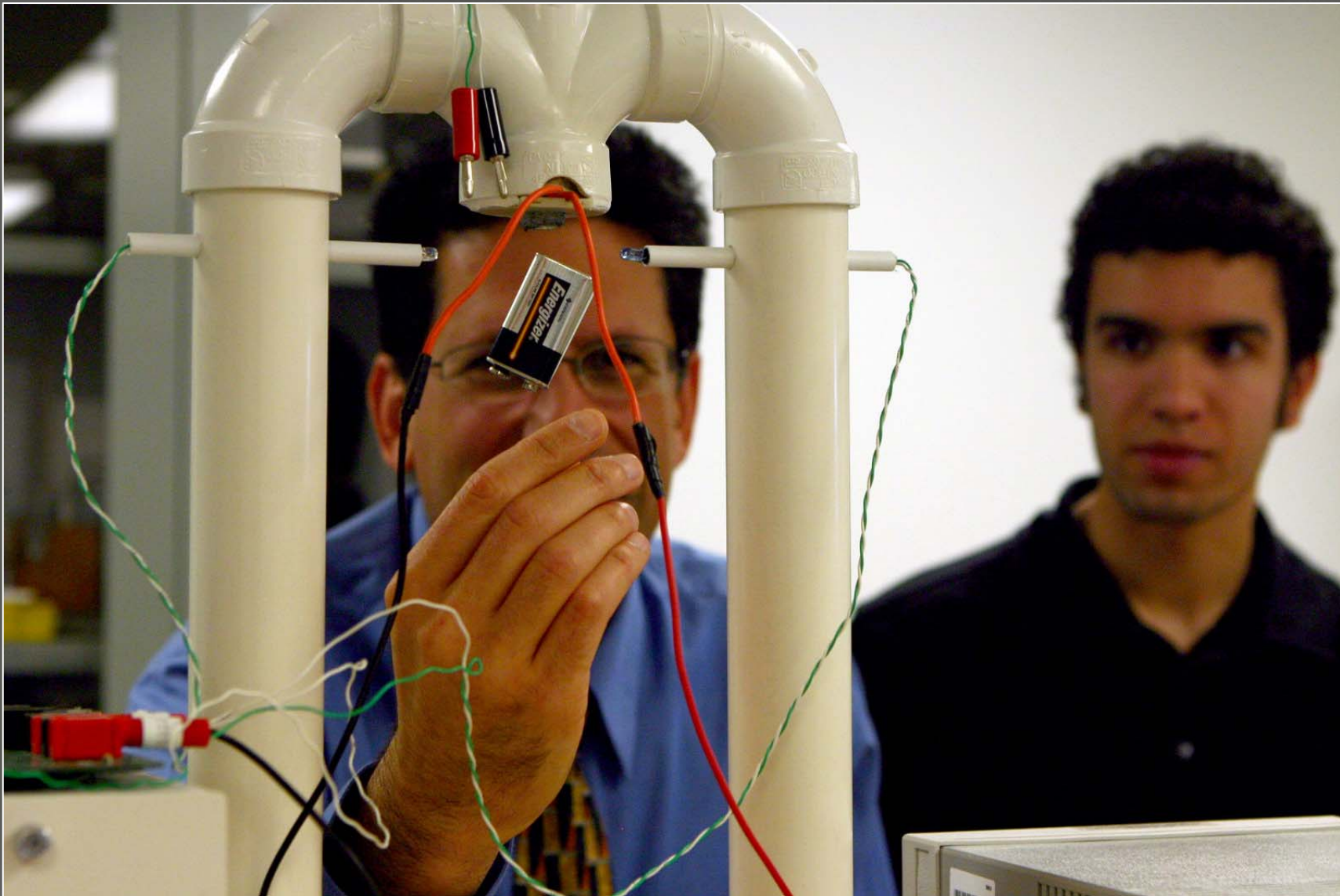
# Levitation of Thin Ring



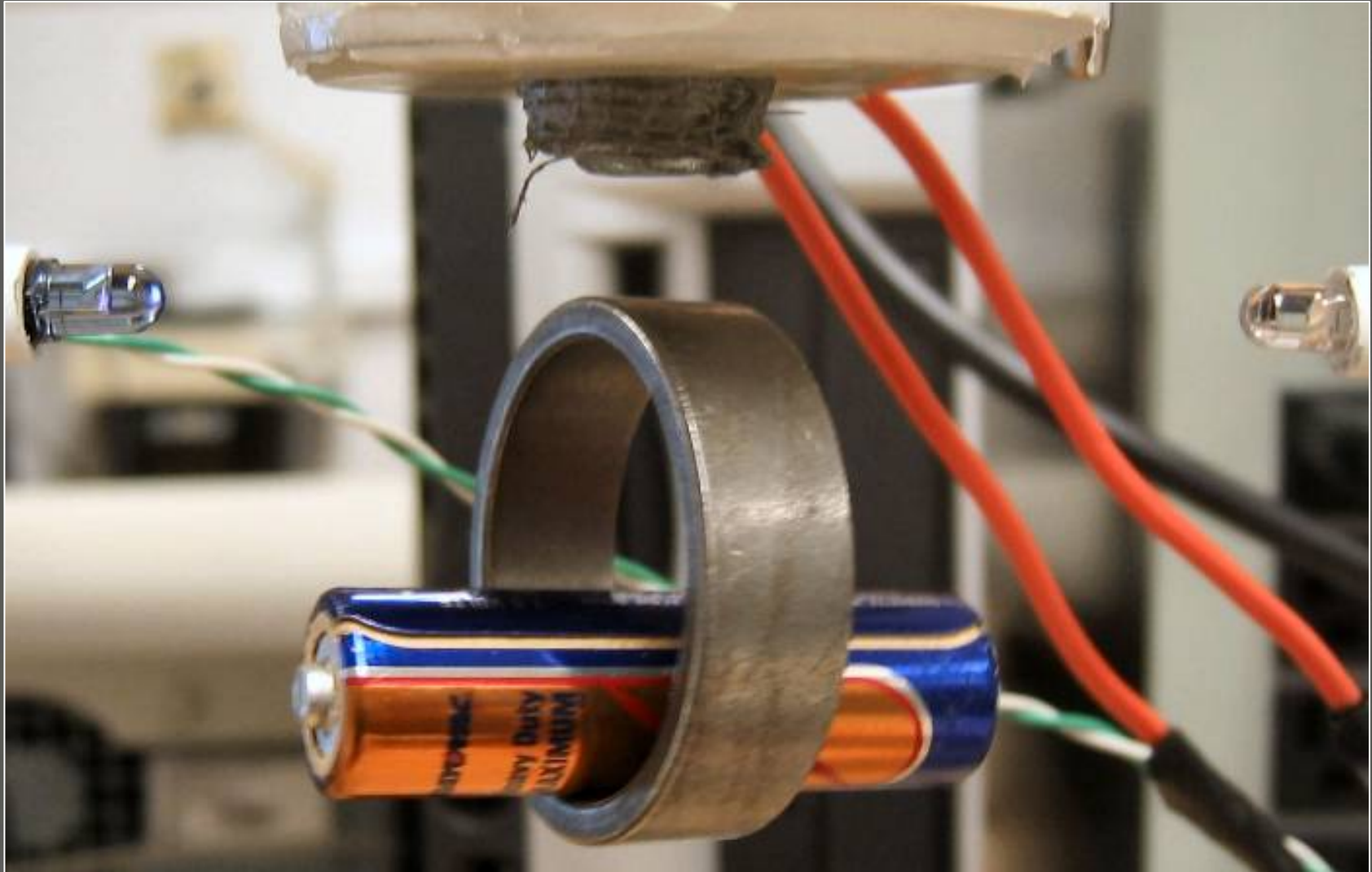
# Levitation of Nested Washer



# Levitation of 9V Battery

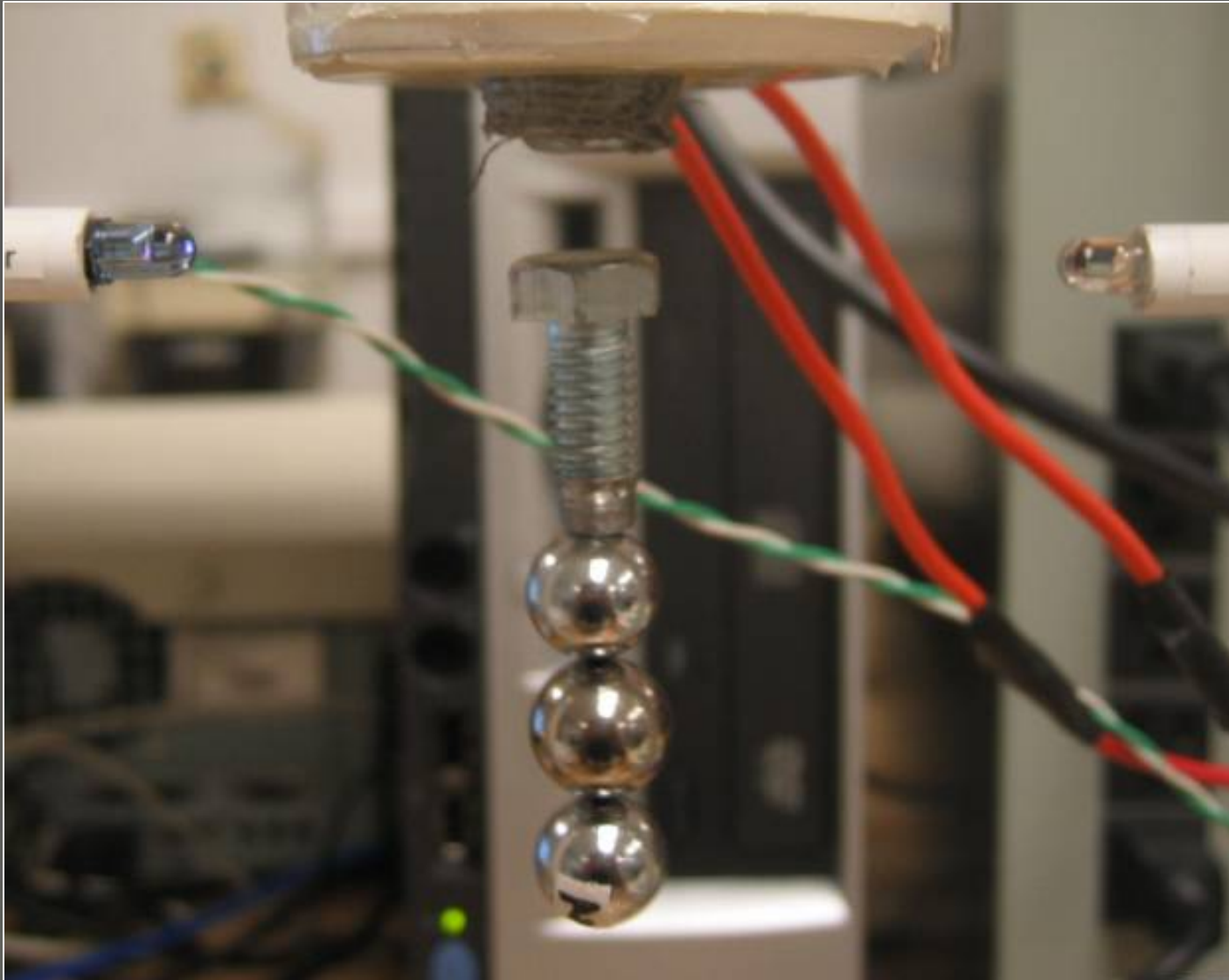


# Levitation of Various Objects

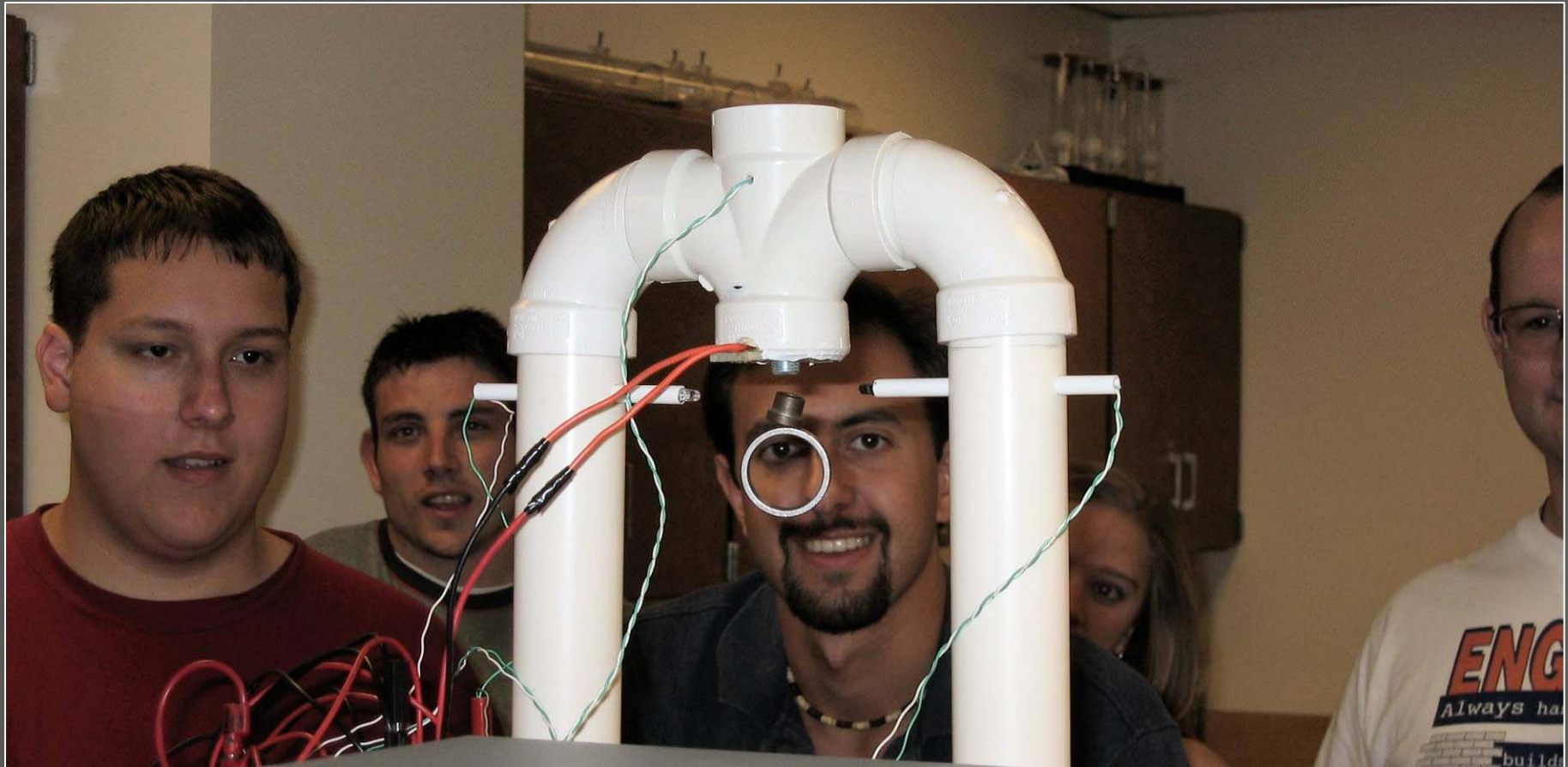




# Levitation of Various Objects



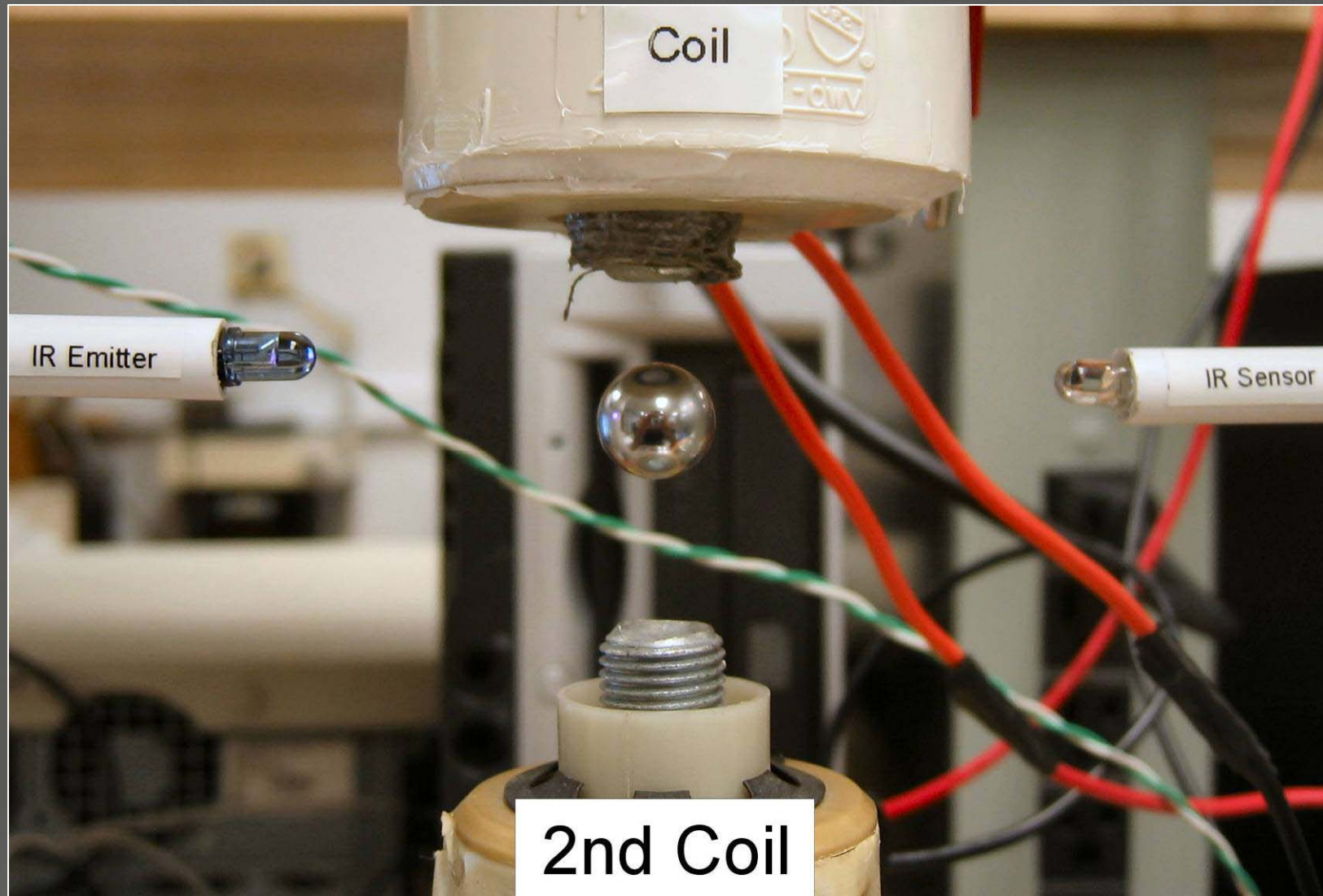
# Levitation of Various Objects



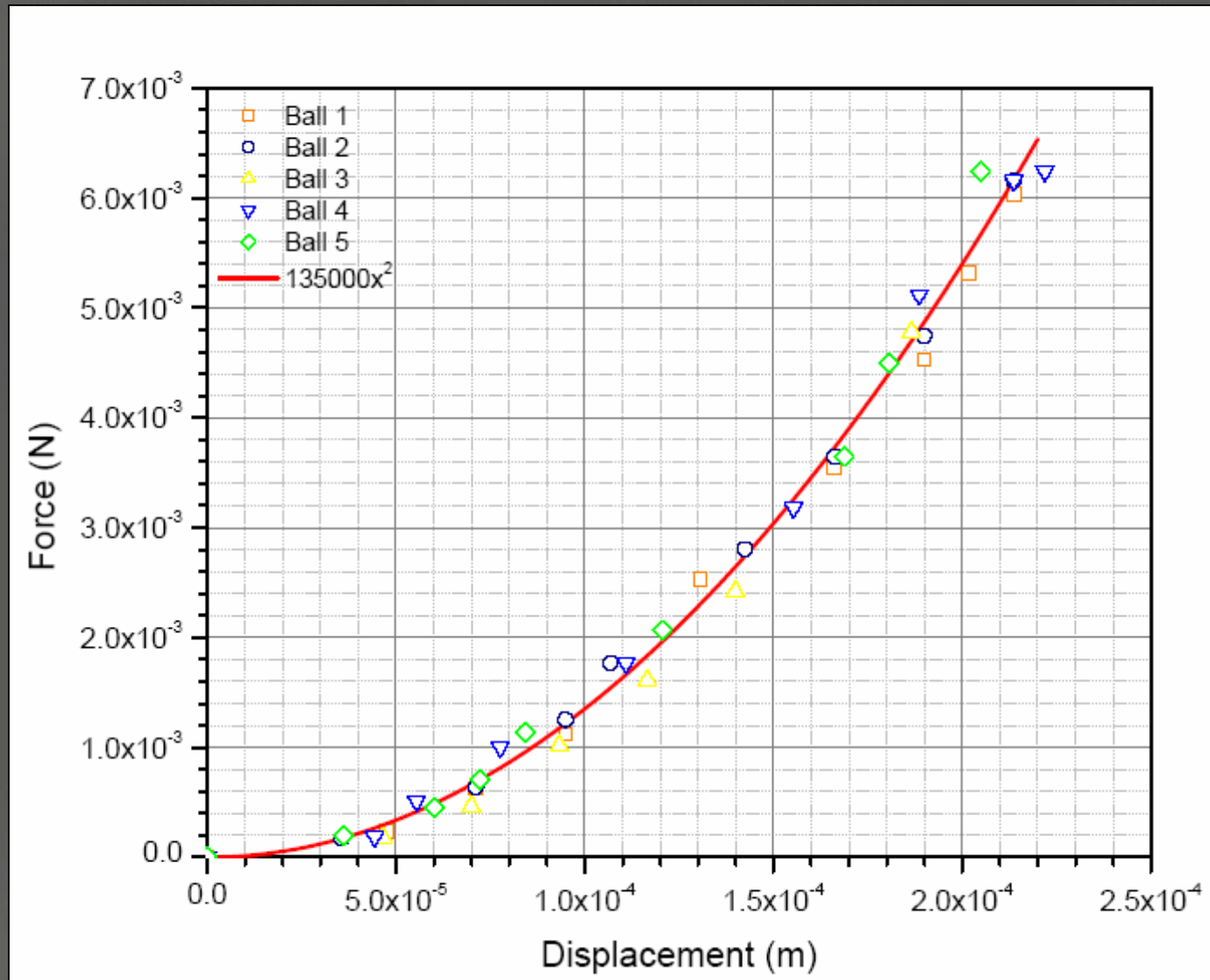
# Current Comparisons

OBJECT	MASS (g)	% Variation From 6.8g	CURRENT (A)				
			A -Lead	A - PD	A - PID	LV - PD	LV - PID
Nested Washer	8.94	29.75%	1.34	1.49	1.47	1.49	1.46
Bolt	10.64	54.43%	1.29	1.36	1.32	1.39	1.37
Bolt 2	14.23	106.53%	1.59	1.68	1.60	1.72	1.72
AA Battery	18.89	174.17%	1.16	1.29	1.24	1.30	1.22
Thin Ring	23.49	240.93%	1.27	1.36	1.34	1.37	1.26
Splined Ring	44.27	542.53%	N/A	1.78	1.79	N/A	1.75
9-Volt Battery	47.36	587.37%	N/A	N/A	2.19	N/A	N/A
Large Bolt	50.95	639.48%	N/A	N/A	2.27	N/A	N/A

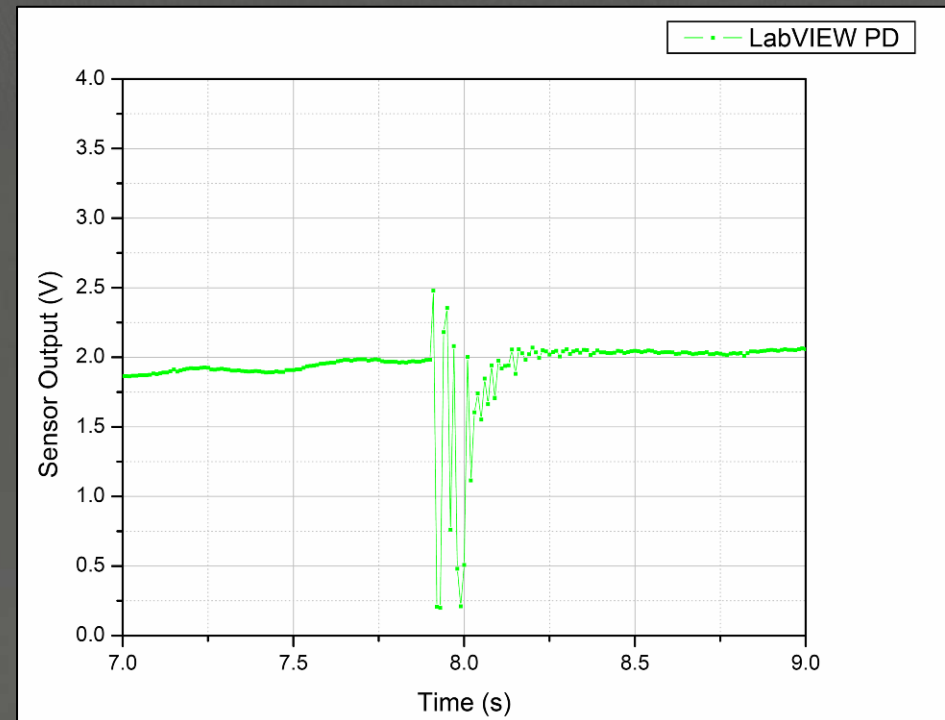
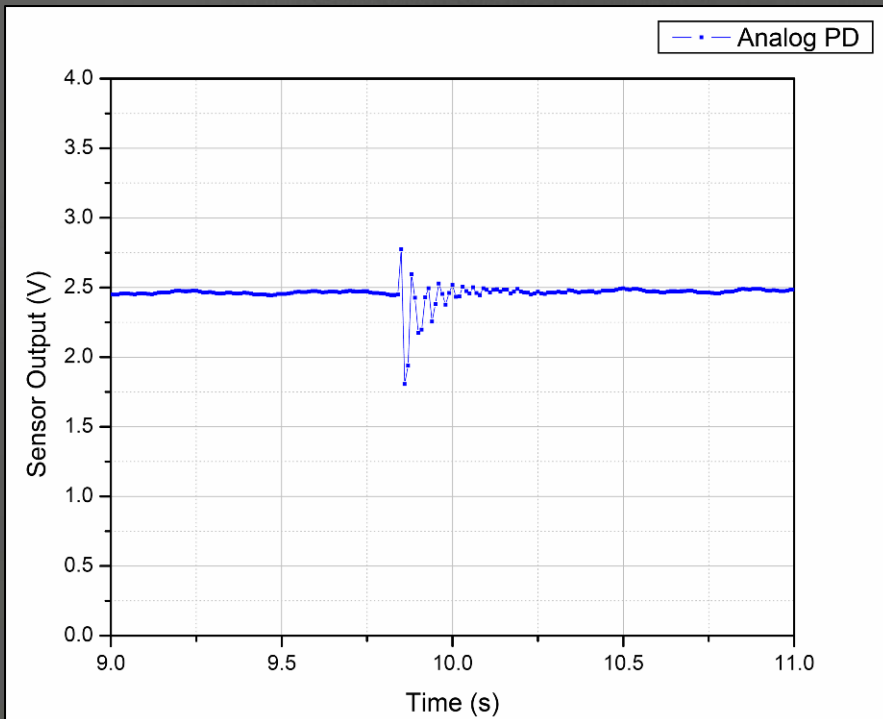
# Addition of Second Coil



# Stiffness



# PD Controller Response



# Outline of Presentation

- o Objectives
- o Background
- o Maglev Testbed
- o Control Strategies
- o Simulation Studies
- o Experimental Studies
- o **Summary**
- o Future Work

# Summary

- o Developed nonlinear and linearized model
- o Conducted simulation studies to compare controllers and select gains
- o Implemented PID-type controllers using analog circuits with NI-ELVIS board
- o Developed software-based controllers using LabVIEW
- o Compared "hardware" vs. "software" based controllers
- o Added second coil for determining stiffness and system response
- o Tested practical differentiator and anti-windup circuits



# Movie



# Closing

- o Applied classical linear control strategies to a linearized model of a nonlinear system
- o Implemented analog controllers using NI-ELVIS
- o Developed LabVIEW controllers
  - o Can dynamically change controller type and parameters and observe effect on system in real-time

# Outline of Presentation

- o Objectives
- o Background
- o Maglev Testbed
- o Control Strategies
- o Simulation Results
- o Experimental Results
- o Summary
- o Future Work**

# Future Work

- o Implement nonlinear and optimal control strategies in LabVIEW & compare performance with linear control strategies
- o Develop a detailed FEA model to optimize the shape of the electromagnet for maximum field strength
- o Design an array of sensors to translate levitated object vertically.
- o Add horizontal electromagnets for two degree-of-freedom system.

# Future Work (cont'd)

- o Add velocity sensor for controlled damping
- o Develop more advanced model
  - o Include gyro-dynamics
  - o Include skin friction using a CFD model
- o Test remote access to VI through internet
- o Create a virtual reality model using Virtual Reality Toolbox
  - o Visualization of simulation without actual implementation

The End

Questions ...

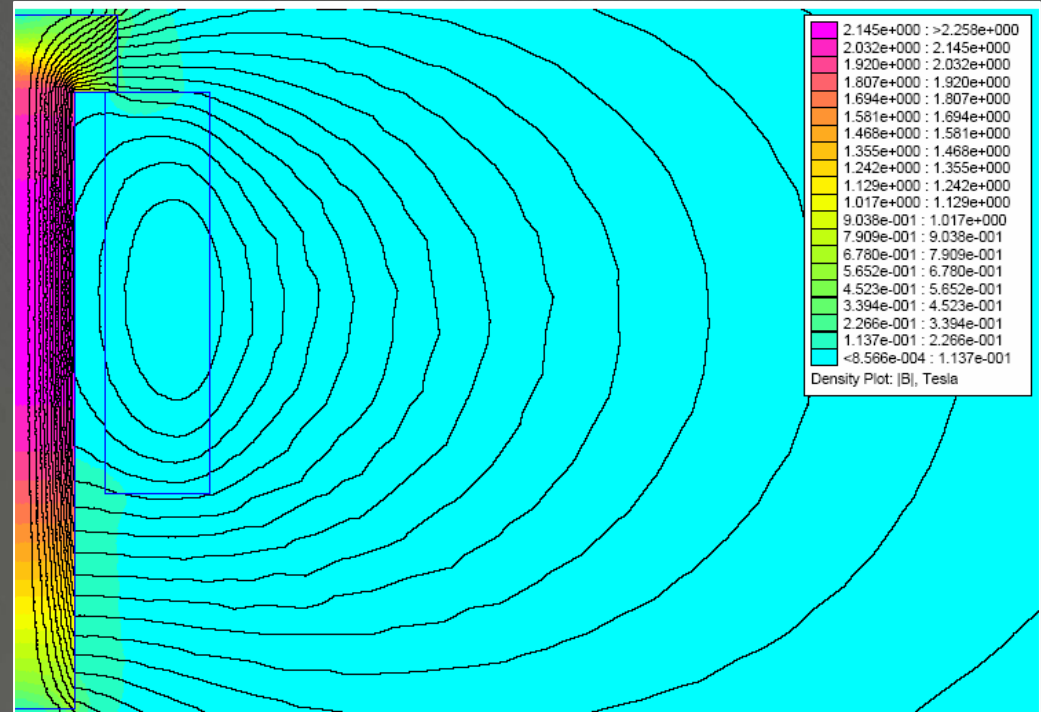
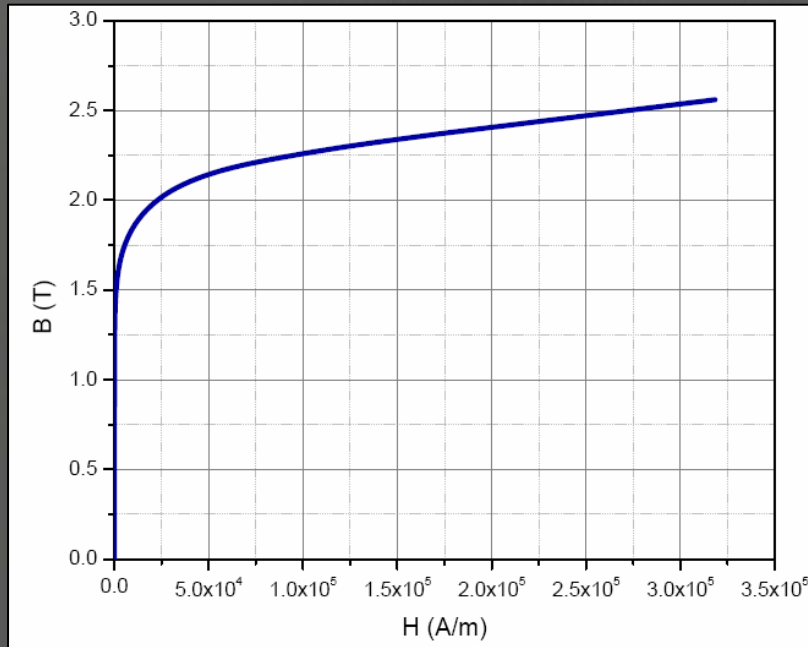
# Movie of Levitated Ball





# Backup Slides

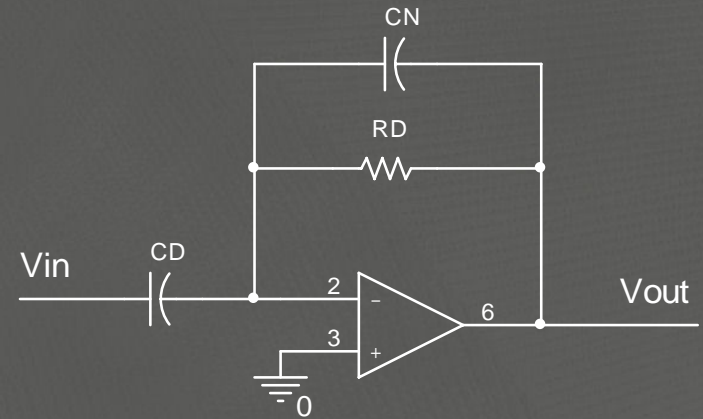
# Magnetic Field FEA Results



Using FEMM package

# Practical Differentiator

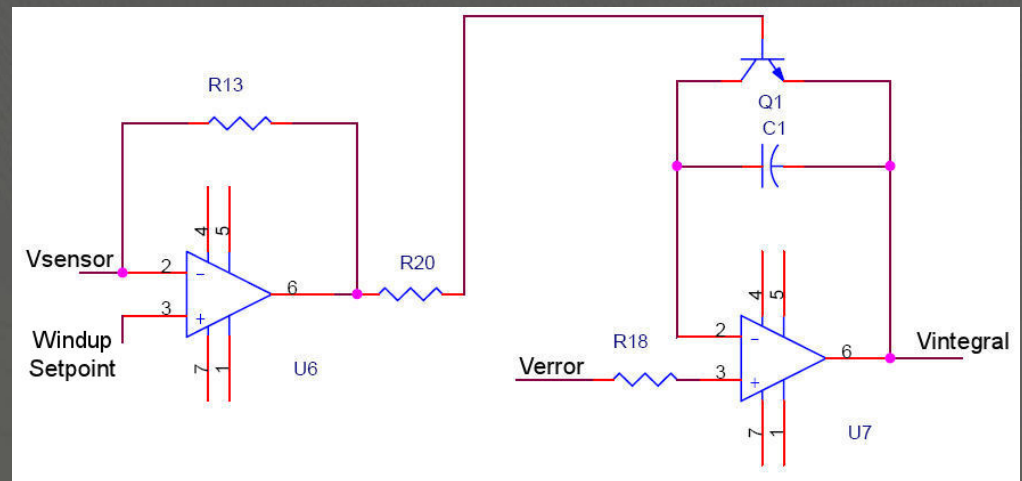
- A pole needs to be added to differentiator circuit for “roll-off” of high frequency response.
- Without additional pole, differentiator circuit just amplifies any high frequency noise signals.
- Need to filter high frequency signals from signal going to the differentiator.
- Capacitor, CN, added in the differentiator circuit



$$\frac{V_{out}(s)}{V_{in}(s)} = \frac{(RD)s}{(RD \times CN)s + 1}$$

# Anti-windup Circuit

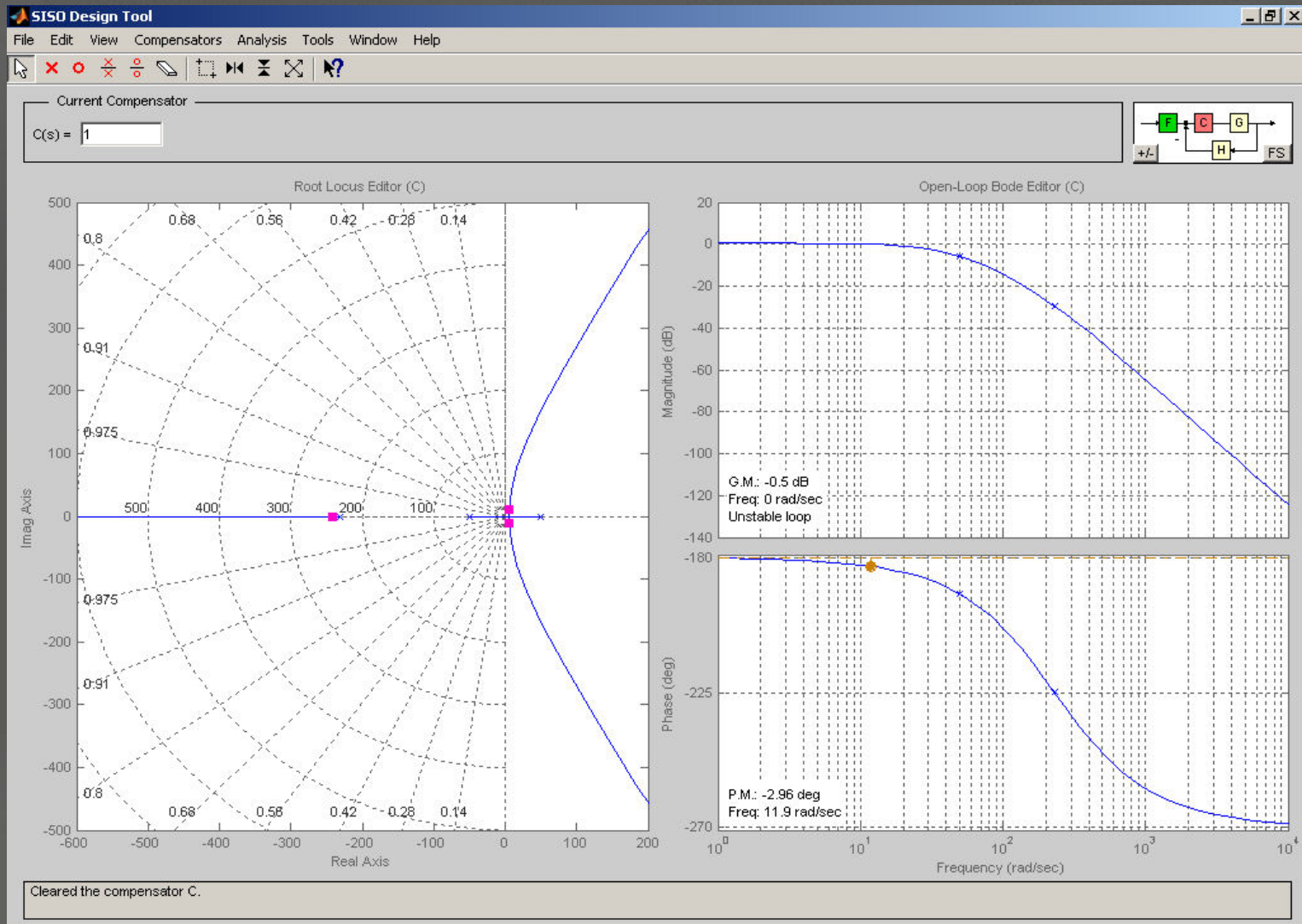
- o The integral circuit integrates the error with respect to time
- o If error is large & constant, e.g., when the object is not present, integral action will continuously increase until saturation
- o Need an “anti-windup” circuit



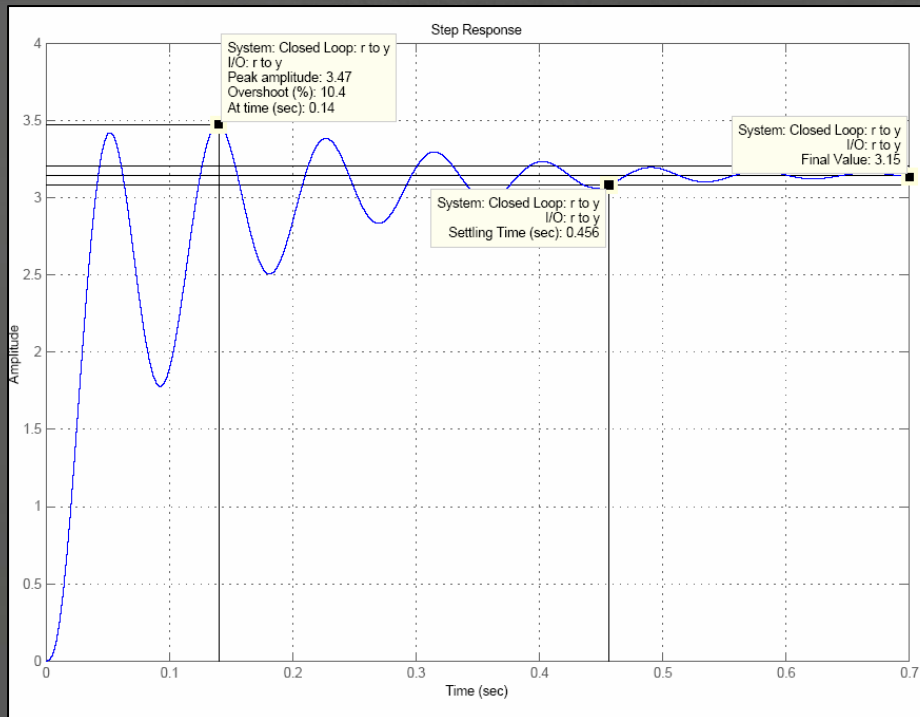
# System Identification

SYMBOL	DESCRIPTION	SI UNITS
$m$	Mass of the object	kg
$x_o$	Equilibrium distance	m
$V_{in}$	Voltage supplied to electromagnet	V
$i_o$	Equilibrium current	A
$C$	Force constant	N-m <sup>2</sup> /A <sup>2</sup>
$R$	Coil resistance	$\Omega$
$L$	Coil inductance	H
$\beta$	Sensor gain	V/m
$\beta_F$	Transistor constant	

# SISO Design Tool



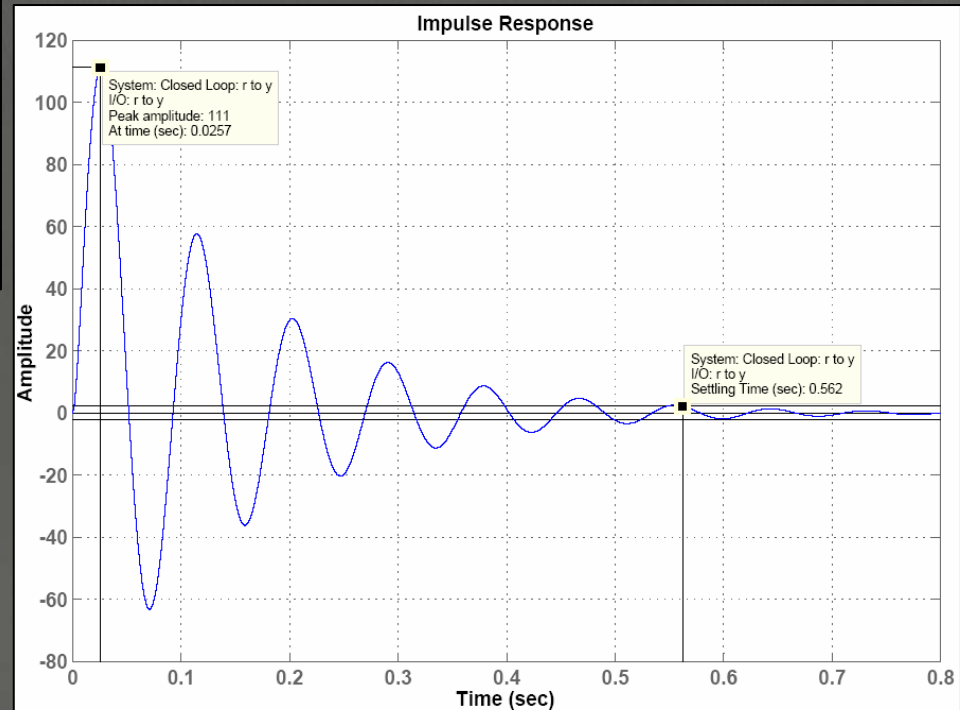
# PD Controller Response



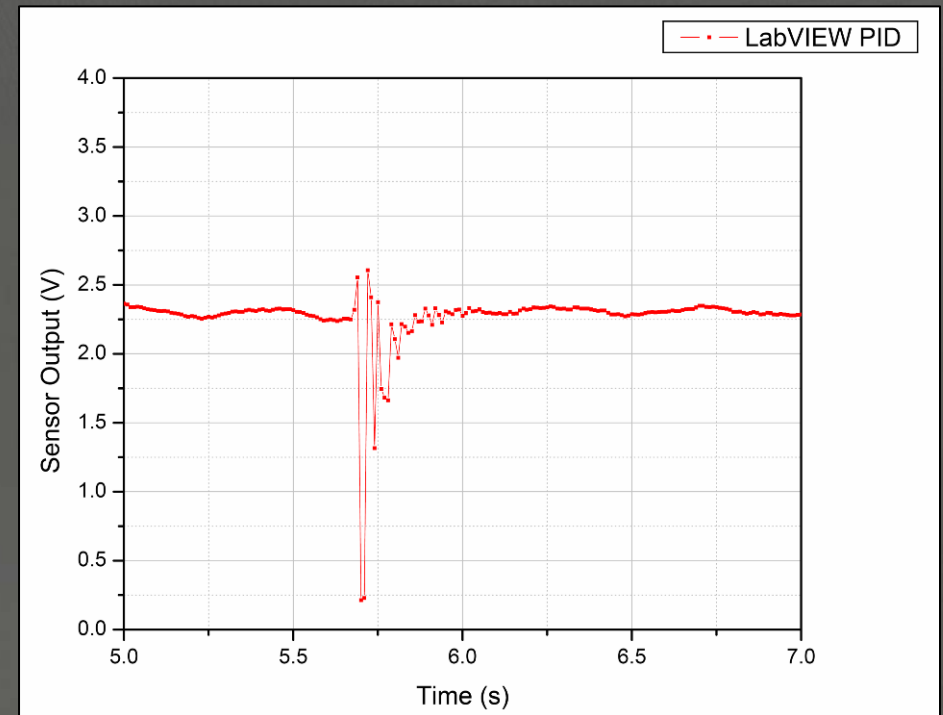
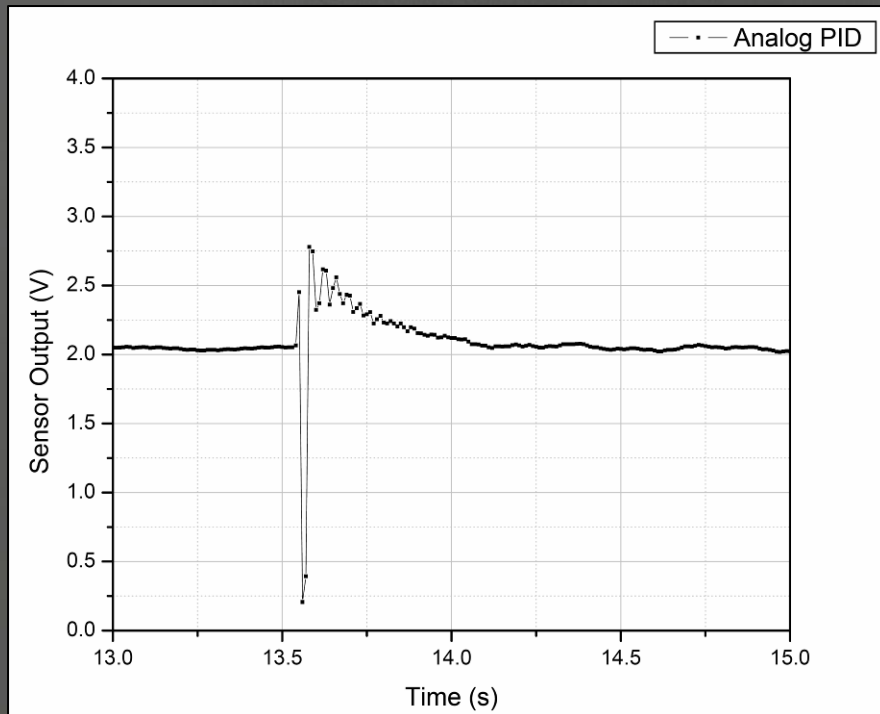
Impulse Response  
Settling Time: 0.562s

Step Response  
Settling Time: 0.157s  
Steady-State Error: 2.15

$$(K_P=0.68, K_D=0.0141)$$

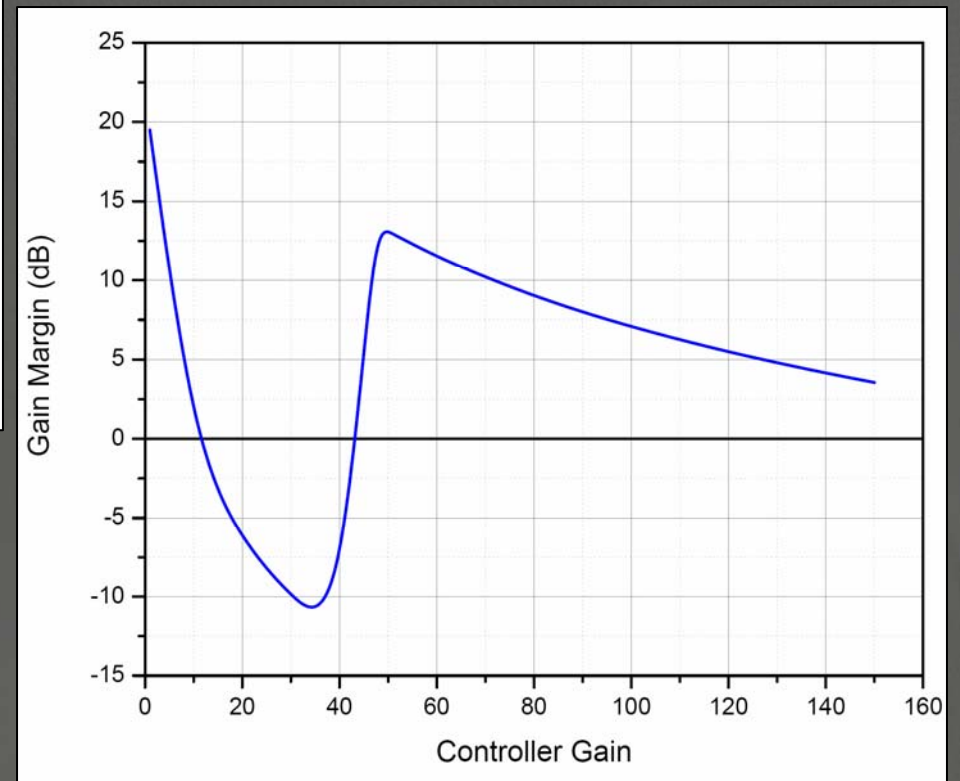
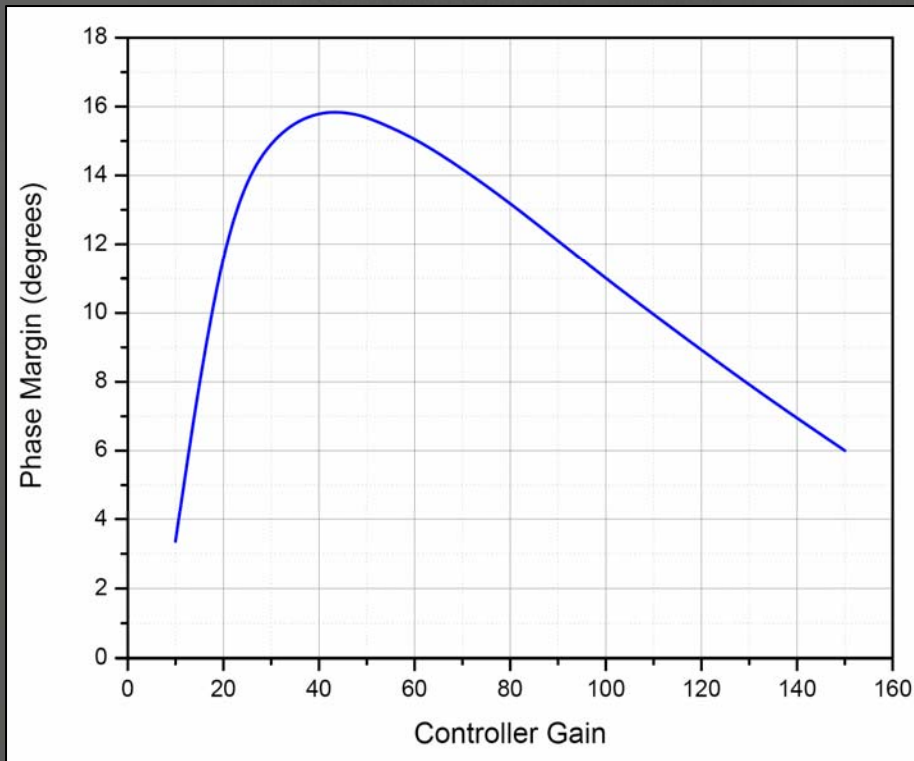


# PID Controller Response





# Margin Plots for PD Controller



# Margin Plots for PID Controller

